

# Introduction to Nonlinear Control

Stability, control design, and estimation

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## Part I:

### Chapter 6: Absolute Stability

6.1 A Commonly Ignored Design Issue

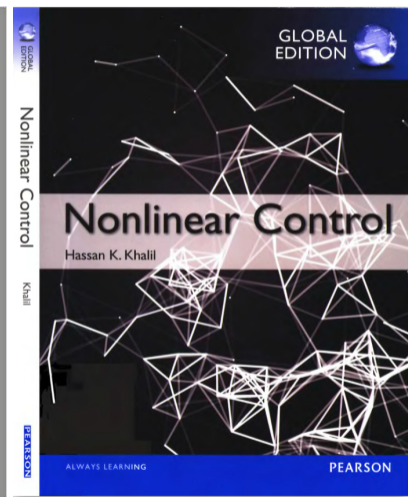
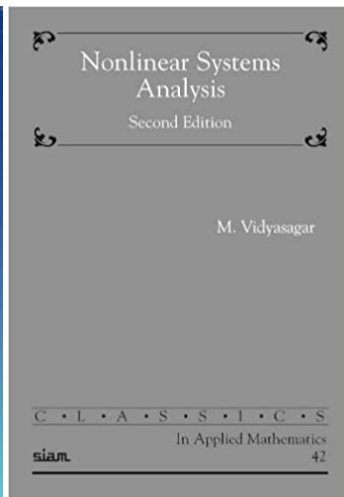
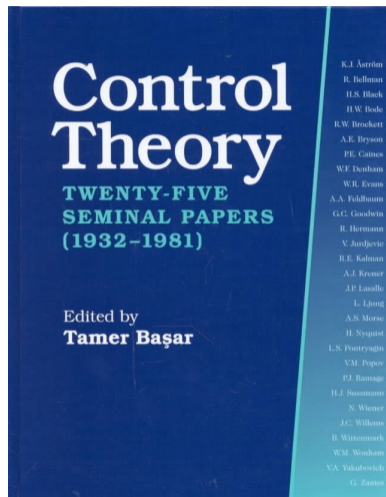
6.2 Historical Perspective on the Lur'e Problem

6.3 Sufficient Conditions for Absolute Stability



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# Absolute Stability



# Absolute Stability

- 1 A Commonly Ignored Design Issue
- 2 Historical Perspective on the Lur'e Problem
- 3 Sufficient Conditions for Absolute Stability
  - Circle Criterion
  - Popov Criterion
  - Circle versus Popov Criterion

## Section 1

### A Commonly Ignored Design Issue

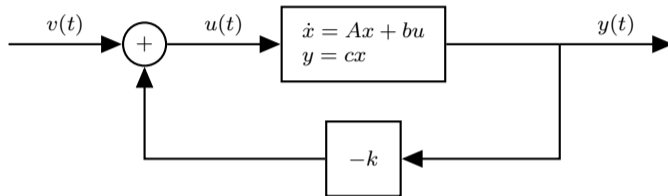
## A Commonly Ignored Design Issue

Linear system:  $(A \in \mathbb{R}^{n \times n}, b \in \mathbb{R}^{n \times 1}, c \in \mathbb{R}^{1 \times n})$

$$\dot{x} = Ax + bu, \quad y = cx,$$

Feedback interconnection:  $u = -ky$

$$\dot{x} = (A - bkc)x,$$



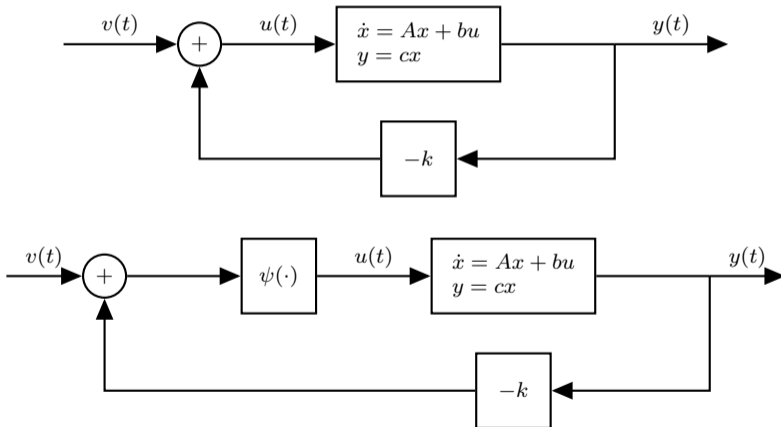
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## A Commonly Ignored Design Issue (2, Example: Pendulum)

### Consider:

- Linearization of the inverted pendulum in the upright position

$$A = \begin{bmatrix} 0 & 1 \\ 9.81 & -0.1 \end{bmatrix}, \quad b = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad c = [ 1 \quad 0 ]$$

(for given parameters)

- We know that the closed loop system  $\dot{x} = A - bkc$ ,  $k = 10$  is asymptotically stable (i.e.,  $A - bkc$  is Hurwitz)

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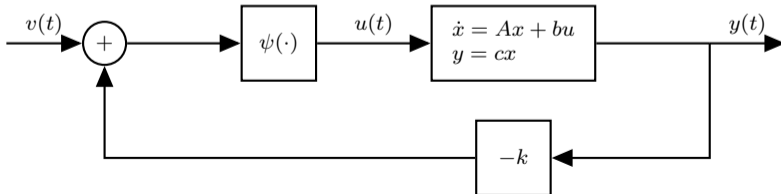
### However:

- Any motor used to drive the cart has limited power  $u \in [u_{lb}, u_{ub}]$ ,  $u_{lb}, u_{ub} \in \mathbb{R}$ .
- Hence,

$$\psi(e) = \begin{cases} u_{lb}, & \text{for } e \leq u_{lb}, \\ e, & \text{for } u_{lb} \leq e \leq u_{ub}, \\ u_{ub}, & \text{for } e \geq u_{ub}, \end{cases}$$

(where  $e(t) = v(t) - ky(t)$  denotes the error variable)

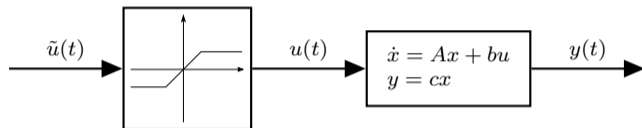
- Question:** Is the origin of  $\dot{x} = Ax - b\psi(kcx)$  asymptotically stable?





## A Commonly Ignored Design Issue (3, Saturations)

Input saturation block diagram:

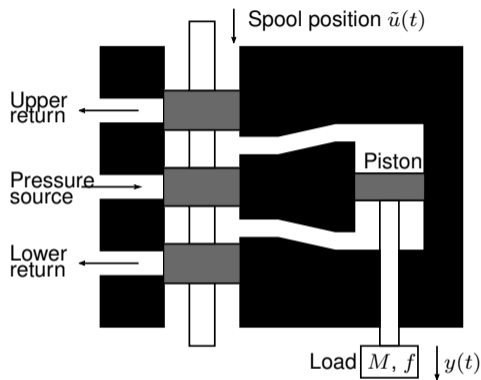


The *saturation function*  $\text{sat} : \mathbb{R} \rightarrow [-1, 1]$ :

$$\text{sat}(y) = \begin{cases} -1, & \text{for } y \leq -1, \\ y, & \text{for } -1 \leq y \leq 1, \\ 1, & \text{for } y \geq 1. \end{cases}$$

- From the normalized function a specific saturation can be obtained through an appropriate scaling and translation.

## A Commonly Ignored Design Issue (4, Example: A servo-valve)



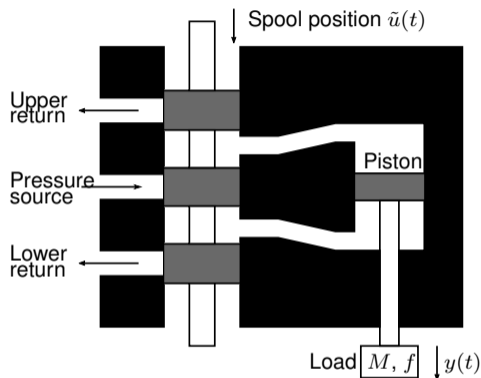
- The linear dynamics are defined through the matrices

$$A = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{B}{M} \end{bmatrix}, \quad b = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad c = \begin{bmatrix} \frac{K}{M} & 0 \end{bmatrix},$$

$$\text{where } K = a \frac{\partial g}{\partial x} \quad \text{and} \quad B = f + \frac{a^2}{\partial P}.$$

- Here,  $g(x, P)$  denotes the flow,  $a$  the area of the piston,  $P$  the pressure, and  $f$  the viscous friction.

## A Commonly Ignored Design Issue (4, Example: A servo-valve)



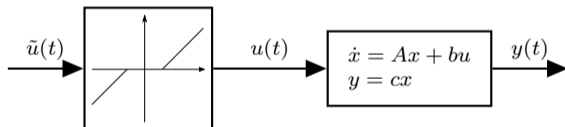
- Raising the spool allows an inflow of pressure
- Simultaneously pressure drop via the upper return so that the piston will rise
- Note the overlap near the openings

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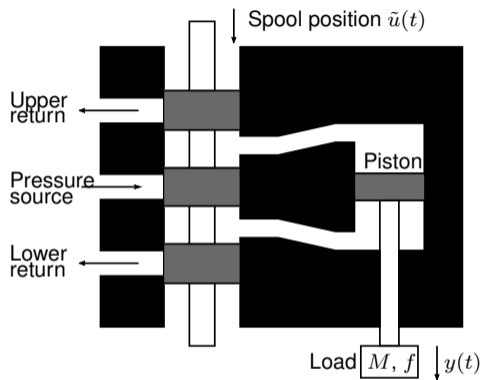
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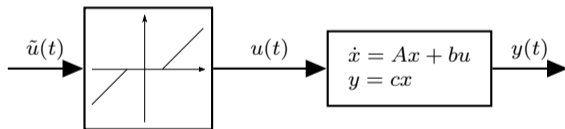
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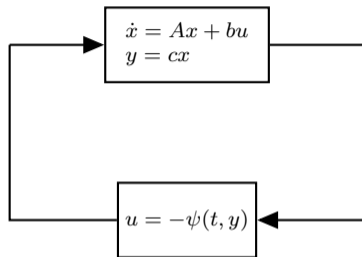


Deadzone  $dz : \mathbb{R} \rightarrow \mathbb{R}$ :

$$dz(y) = \begin{cases} y + 1, & \text{for } y \leq -1, \\ 0, & \text{for } -1 \leq y \leq 1, \\ y - 1, & \text{for } y \geq 1. \end{cases}$$

## A Commonly Ignored Design Issue (5, The Lur'e Problem)

Consider the feedback interconnection:



Lur'e problem:

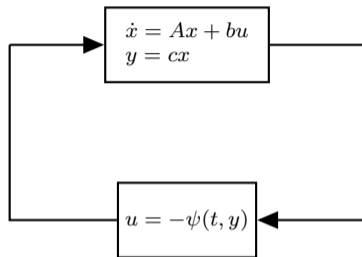
- Which conditions on the functions  $\psi : \mathbb{R}_{\geq 0} \times \mathbb{R} \rightarrow \mathbb{R}$  guarantee asymptotic stability of the origin?

Note that:

- The nonlinearity can be time-dependent
- We assume that the reference signal  $v(t)$  is zero.
- While we focus on the SISO case, many results can be extended to the MIMO case.

## A Commonly Ignored Design Issue (5, The Lur'e Problem)

Consider the feedback interconnection:



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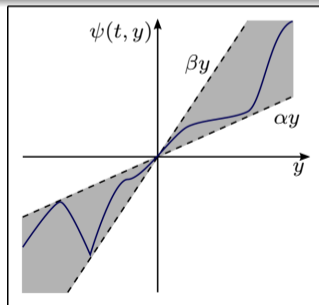
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### Definition (Sector condition)

Let  $\alpha, \beta \in \mathbb{R}$ ,  $\alpha < \beta$ , and  $\Omega \subset \mathbb{R}$ . A nonlinearity  $\psi : \mathbb{R}_{\geq 0} \times \mathbb{R} \rightarrow \mathbb{R}$  satisfies a sector condition if

$$\alpha y^2 \leq y\psi(t, y) \leq \beta y^2$$

for all  $t \geq 0$  and for all  $y \in \Omega$ . For  $\Omega = \mathbb{R}$  we say that the sector condition is satisfied globally.



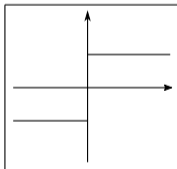
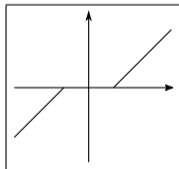
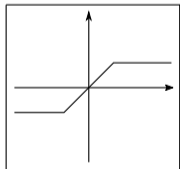
# A Commonly Ignored Design Issue (6, The Sector Condition)

Common nonlinearities:  $\text{sign} : \mathbb{R} \rightarrow \mathbb{R}$ ,

$$\text{sat}(y) = \begin{cases} -1, & \text{for } y \leq -1, \\ y, & \text{for } -1 \leq y \leq 1, \\ 1, & \text{for } y \geq 1. \end{cases}$$

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$$\text{sign}(y) = \begin{cases} -1, & \text{for } y < 0, \\ 0, & \text{for } y = 0, \\ 1, & \text{for } y > 0, \end{cases}$$



Question:

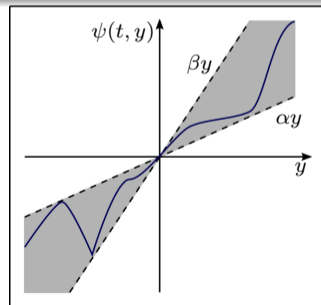
- Which nonlinearity satisfies a sector condition?

## Definition (Sector condition)

Let  $\alpha, \beta \in \mathbb{R}$ ,  $\alpha < \beta$ , and  $\Omega \subset \mathbb{R}$ . A nonlinearity  $\psi : \mathbb{R}_{\geq 0} \times \mathbb{R} \rightarrow \mathbb{R}$  satisfies a sector condition if

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## A Commonly Ignored Design Issue (7, Absolute Stability)

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### Definition (Absolute stability)

Let  $\alpha, \beta \in \mathbb{R}$ ,  $\alpha < \beta$ , and  $\Omega \subset \mathbb{R}$ . The Lur'e system

$$\dot{x} = Ax - b\psi(t, y)$$

is called **absolutely stable** (with respect to  $\alpha, \beta, \Omega$ ) if the origin is asymptotically stable for all  $\psi : \mathbb{R}_{\geq 0} \times \mathbb{R} \rightarrow \mathbb{R}$  satisfying the sector condition for all  $t \geq 0$  and for all  $y_0 \in \Omega$ .



# Historical Perspective on the Lur'e Problem

## Conjecture (Aizerman's Conjecture (1949))

Let  $\alpha, \beta \in \mathbb{R}$ ,  $\alpha < \beta$ , and suppose the origin of the linear system  $\dot{x} = Ax + bu$ ,  $y = cx$  is globally asymptotically stable for all linear feedbacks

$$u = -\psi(y) = -ky, \quad k \in [\alpha, \beta].$$

Then the origin is globally asymptotically stable for all nonlinear feedbacks in the sector

$$\alpha \leq \frac{\psi(y)}{y} \leq \beta, \quad y \neq 0.$$

↪ Conjecture was shown to be wrong through counterexamples.

## Conjecture (Kalman's Conjecture (1957))

Let  $\alpha, \beta \in \mathbb{R}$ ,  $\alpha < \beta$ , and suppose the origin of the linear system  $\dot{x} = Ax + bu$ ,  $y = cx$  is globally asymptotically stable for all linear feedbacks

$$u = -\psi(y) = -ky, \quad k \in [\alpha, \beta].$$

Then the origin is globally asymptotically stable for all nonlinear feedbacks belonging to the incremental sector

$$\alpha \leq \frac{\partial}{\partial y} \psi(y) \leq \beta.$$

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## Section 2

### Historical Perspective on the Lur'e Problem

## Historical Perspective on the Lur'e Problem (Detour into Frequency domain)

### Definition (Positive real)

A transfer function  $H(s)$  is positive real if

$$\operatorname{Re}(H(s)) \geq 0 \quad \text{for all } s \in \overline{\mathbb{C}}_+.$$

The transfer function is strictly positive real (SPR) if  $H(s - \varepsilon)$  is positive real for some  $\varepsilon > 0$ .

Note that the strictly positive real definition above is equivalent to the requirement that

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### Example

Consider  $H(s) = \frac{1}{s}$ . Then

$$\operatorname{Re}(H(s)) = \operatorname{Re}\left(\frac{1}{\sigma + j\omega}\right) = \operatorname{Re}\left(\frac{\sigma - j\omega}{\sigma^2 + \omega^2}\right) = \frac{\sigma}{\sigma^2 + \omega^2} \geq 0$$

for  $\sigma \geq 0$ . So  $H(s)$  is positive real.

Consider  $H(s) = \frac{1}{s+1}$ . Then

$$\operatorname{Re}(H(s)) = \operatorname{Re}\left(\frac{1}{\sigma + j\omega + 1}\right) = \operatorname{Re}\left(\frac{\sigma + 1 - j\omega}{(\sigma + 1)^2 + \omega^2}\right) = \frac{\sigma + 1}{\sigma^2 + \omega^2} > 0$$

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for  $\sigma \geq 0$ . So  $H(s)$  is strictly positive real.

## Lemma

The transfer function  $H(s)$  is strictly positive real if and only if  $H(s)$  is Hurwitz (i.e., has all its poles in the open left-half complex plane) and

$$\operatorname{Re}(H(j\omega)) > 0, \quad \text{for all } \omega \in \mathbb{R}.$$

## Lemma (Kalman-Yakubovich-Popov Lemma)

Let

$$H(s) = c(sI - A)^{-1}b + d$$

be a transfer function where  $A \in \mathbb{R}^{n \times n}$  is Hurwitz,  $(A, b)$  is controllable, and  $(A, c)$  is observable. Then  $H(s)$  is strictly positive real if and only if there exist  $P \in \mathcal{S}_{>0}^n$ , a (row) vector  $L \in \mathbb{R}^{1 \times n}$ , a number  $w \in \mathbb{R}$ , and positive constant  $\varepsilon > 0$  such that

$$A^T P + PA = -L^T L - \varepsilon P,$$

$$Pb = c^T - L^T w,$$

$$w^2 = 2d.$$

## Section 3

### Sufficient Conditions for Absolute Stability

## Sufficient Conditions for Absolute Stability

### Theorem (Absolute stability)

Assume that  $A$  is Hurwitz,  $(A, b, c, d)$  is controllable and observable,  $H(s) = c(sI - A)^{-1}b + d$  is strictly positive real, and  $\psi(t, y)$  is in the sector  $[0, \infty)$ . Then the origin of the Lur'e system  $\dot{x} = Ax - b\psi(t, y)$  is globally exponentially stable (i.e., the system is *absolutely stable*).

## Sufficient Conditions for Absolute Stability

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- Result is only applicable to systems with  $A$  Hurwitz
- **Idea:** apply loop transformation to generalize the result



# Sufficient Conditions for Absolute Stability

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Note that:

- Result is only applicable to systems with  $A$  Hurwitz
- **Idea:** apply loop transformation to generalize the result

If  $\psi(t, y)$  satisfies

$$\alpha y^2 \leq y\psi(t, y) \leq \beta y^2, \quad \alpha, \beta \in \mathbb{R}$$

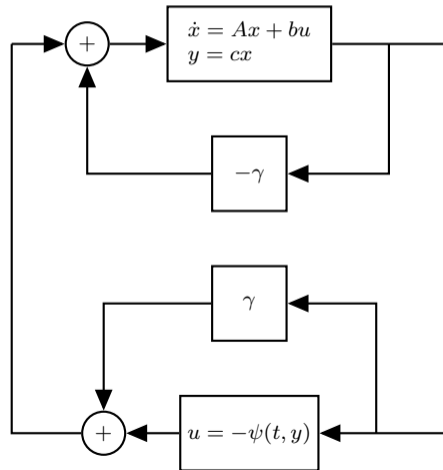
then  $\hat{\psi}(t, y) = \psi(t, y) - \gamma y$  satisfies the sector condition

$$\hat{\alpha} y^2 \leq y\hat{\psi}(t, y) \leq \hat{\beta} y^2 \quad \hat{\alpha} = \alpha - \gamma, \hat{\beta} = \beta - \gamma, \quad \gamma \in \mathbb{R}$$

Let  $\gamma \in \mathbb{R}$  such that  $\hat{A} = A - \gamma bc$  is Hurwitz.

The loop transformation satisfies the assumptions of the Theorem as long as  $\hat{\alpha} > 0$ .

$$\begin{aligned} \dot{x} &= Ax + b(-\gamma cx + \gamma cx - \psi(t, y)) \\ &= (A - \gamma bc)x - b(\psi(t, y) - \gamma y) \\ &= \hat{A}x - b\hat{\psi}(t, y) \end{aligned}$$



## Circle Criterion

Consider the Lur'e system  $\dot{x} = Ax - b\psi(t, y)$  with

$$\alpha y^2 \leq y\psi(t, y) \leq \beta y^2, \quad \alpha, \beta \in \mathbb{R}$$

Consider the loop transformation

$$\begin{aligned}\dot{x} &= Ax + b(-\gamma cx + \gamma cx - \psi(t, y)) \\ &= (A - \gamma bc) - b(\psi(t, y) - \gamma y) = \hat{A}x - b\hat{\psi}(t, y)\end{aligned}$$

with  $0 \leq y\hat{\psi}(t, y) \leq \hat{\beta}y^2$ ,  $\gamma = \alpha$ ,  $\hat{\beta} \doteq \beta - \alpha$

### Lemma

Consider the Lur'e system with  $(A, b, c)$  controllable and observable, and transfer function  $G(s) = c(sI - A)^{-1}b$ . The system  $(A, b, c)$  is absolutely stable (with respect to  $\alpha, \beta \in \mathbb{R}, \alpha < \beta, \Omega = \mathbb{R}$ ) if  $\hat{A} = A - \alpha bc$  is Hurwitz, and

$$H(s) = \frac{1 + \beta G(s)}{1 + \alpha G(s)}$$

is strictly positive real; i.e., if

$$\operatorname{Re}(H(s)) = \operatorname{Re}\left(\frac{1 + \beta G(s)}{1 + \alpha G(s)}\right) > 0 \quad \text{for all } s \in \overline{\mathbb{C}}_+.$$

## Circle Criterion (2)

### Definitions: (Disc in the complex plane)

- center  $\sigma : \mathbb{R} \setminus \{0\} \times \mathbb{R}_{>0} \rightarrow \mathbb{R}$
- radius  $r : \mathbb{R} \setminus \{0\} \times \mathbb{R}_{>0} \rightarrow \mathbb{R}$
- for  $\alpha \neq 0$  and  $\beta > 0$  we define

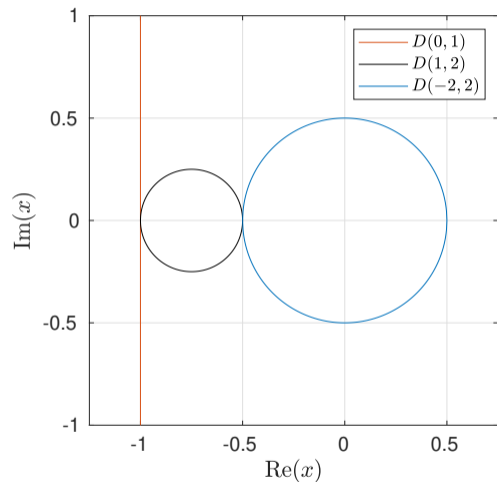
$$\sigma(\alpha, \beta) = \frac{1}{2} \left( \frac{1}{\alpha} + \frac{1}{\beta} \right), \quad r(\alpha, \beta) = \frac{\text{sign}(\alpha)}{2} \left( \frac{1}{\alpha} - \frac{1}{\beta} \right)$$

Then, the disc  $D(\cdot, \cdot)$  is defined as

$$D(\alpha, \beta) = \begin{cases} \{x \in \mathbb{C} : x = -\frac{1}{\beta} + j\omega, \omega \in \mathbb{R}\}, & \text{if } \alpha = 0 < \beta, \\ \{x \in \mathbb{C} : |x - \sigma(\alpha, \beta)| = r(\alpha, \beta)\}, & \text{if } 0 < \alpha < \beta, \\ \{x \in \mathbb{C} : |x - \sigma(\alpha, \beta)| = r(\alpha, \beta)\}, & \text{if } \alpha < 0 < \beta. \end{cases}$$

### Note that

- for  $\alpha \neq 0$ ,  $D(\alpha, \beta)$  defines a disc centered around  $\sigma(\alpha, \beta)$  with radius  $r(\alpha, \beta)$
- for  $\alpha = 0$ ,  $D(0, \beta)$  defines a vertical line



## Circle Criterion (3), (First Case)

First case ( $\alpha = 0 < \beta$ ):  $H(s)$  reduces to

$$H(s) = 1 + \beta G(s)$$

Then

- $H(s)$  has the same poles as  $G(s)$

According to the Lemma

- $G(s)$  (or  $A$ ) needs to be Hurwitz
- $\operatorname{Re}(1 + \beta G(j\omega)) > 0 \quad \forall \omega \in \mathbb{R}$  needs to be satisfied

Note that

- since  $\beta > 0$ , the last item is equivalent to

$$\operatorname{Re}(G(j\omega)) > -\frac{1}{\beta}, \quad \text{for all } \omega \in \mathbb{R}.$$

- this corresponds to the Nyquist plot of  $G(j\omega)$  being to the right of the vertical line in the complex plane through  $-1/\beta$ .

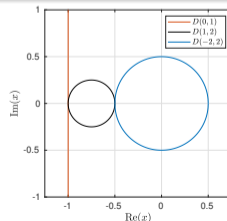
### Lemma

Consider the Lur'e system with  $(A, b, c)$  controllable and observable, and transfer function  $G(s) = c(sI - A)^{-1}b$ . The system  $(A, b, c)$  is absolutely stable (with respect to  $\alpha, \beta \in \mathbb{R}, \alpha < \beta, \Omega = \mathbb{R}$ ) if  $\hat{A} = A - \alpha bc$  is Hurwitz, and

$$H(s) = \frac{1 + \beta G(s)}{1 + \alpha G(s)}$$

is strictly positive real; i.e., if

$$\operatorname{Re}(H(s)) = \operatorname{Re} \left( \frac{1 + \beta G(s)}{1 + \alpha G(s)} \right) > 0 \quad \text{for all } s \in \bar{\mathbb{C}}_+.$$



## Circle Criterion (3), (First Case)

First case ( $\alpha = 0 < \beta$ ):  $H(s)$  reduces to

$$H(s) = 1 + \beta G(s)$$

Then

- $H(s)$  has the same poles as  $G(s)$

According to the Lemma

- $G(s)$  (or  $A$ ) needs to be Hurwitz
- $\operatorname{Re}(1 + \beta G(j\omega)) > 0 \quad \forall \omega \in \mathbb{R}$  needs to be satisfied

Note that

- since  $\beta > 0$ , the last item is equivalent to

$$\operatorname{Re}(G(j\omega)) > -\frac{1}{\beta}, \quad \text{for all } \omega \in \mathbb{R}.$$

- this corresponds to the Nyquist plot of  $G(j\omega)$  being to the right of the vertical line in the complex plane through  $-1/\beta$ .

Recall the disc  $D(\cdot, \cdot)$  for  $\alpha = 0 < \beta$ :

$$D(\alpha, \beta) = \{x \in \mathbb{C} : x = -\frac{1}{\beta} + j\omega, \omega \in \mathbb{R}\}$$

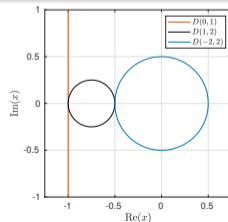
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## Circle Criterion (4), (Second Case)

**Second case** ( $0 < \alpha < \beta$ ):  $H(s)$  is strictly positive real if

$$\operatorname{Re}(H(s)) = \operatorname{Re} \left( \frac{1 + \beta G(j\omega)}{1 + \alpha G(j\omega)} \right) > 0 \quad \text{for all } \omega \in \mathbb{R}$$

Let  $q = G(j\hat{\omega})$  and choose the polar coordinates

$r_1, r_2 > 0$  and  $\theta_1, \theta_2 \in (-\pi, \pi)$  so that

$$\frac{1}{\beta} + q = r_1 e^{j\theta_1} \quad \text{and} \quad \frac{1}{\alpha} + q = r_2 e^{j\theta_2}.$$

Then we can rewrite

$$\begin{aligned} 0 < \operatorname{Re} \left( \frac{\frac{1}{\beta} + G(j\omega)}{\frac{1}{\alpha} + G(j\omega)} \right) &= \operatorname{Re} \left( \frac{r_1 e^{j\theta_1}}{r_2 e^{j\theta_2}} \right) = \operatorname{Re} \left( \frac{r_1}{r_2} e^{j(\theta_1 - \theta_2)} \right) \\ &= \frac{r_1}{r_2} \operatorname{Re}(\cos(\theta_1 - \theta_2) + j \sin(\theta_1 - \theta_2)) = \frac{r_1}{r_2} \cos(\theta_1 - \theta_2) \end{aligned}$$

$$\rightsquigarrow |\theta_1 - \theta_2| < \frac{\pi}{2}$$

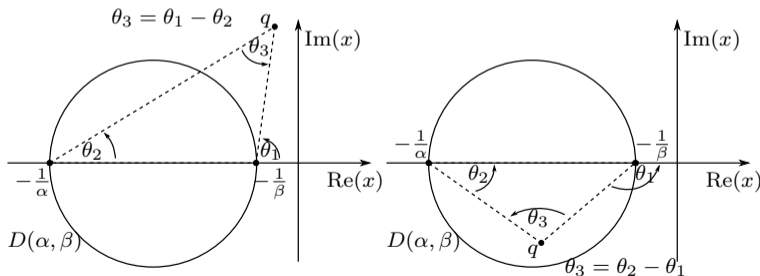
$\rightsquigarrow \operatorname{Re}(H(s)) > 0 \quad \forall \omega \in \mathbb{R} \rightsquigarrow$  Nyquist plot lies entirely outside the disc.

**Note that:**

- $H(s)$  needs to be Hurwitz
- Poles:  $0 = 1 + \alpha G(j\omega)$
- the Nyquist plot needs to encircle the point  $-1/\alpha$  as many times as there are right-half plane poles of  $G(s)$ .

**Absolute stability:**

The Nyquist plot does not enter the disc and encircles the disc as many times as there are right-half plane poles of  $G(s)$ .



## Circle Criterion (5), (Third Case)

Third case ( $\alpha < 0 < \beta$ ):  $H(s)$  is strictly positive real if

$$\operatorname{Re}(H(s)) = \operatorname{Re}\left(\frac{1 + \beta G(j\omega)}{1 + \alpha G(j\omega)}\right) > 0 \quad \text{for all } \omega \in \mathbb{R}$$

Let  $q = G(j\omega)$  and choose the polar coordinates

$r_1, r_2 > 0$  and  $\theta_1, \theta_2 \in (-\pi, \pi)$  so that

$$\frac{1}{\beta} + q = r_1 e^{j\theta_1} \quad \text{and} \quad \frac{1}{\alpha} + q = r_2 e^{j\theta_2}.$$

Then we can rewrite

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$$\rightsquigarrow |\theta_1 - \theta_2| > \frac{\pi}{2}$$

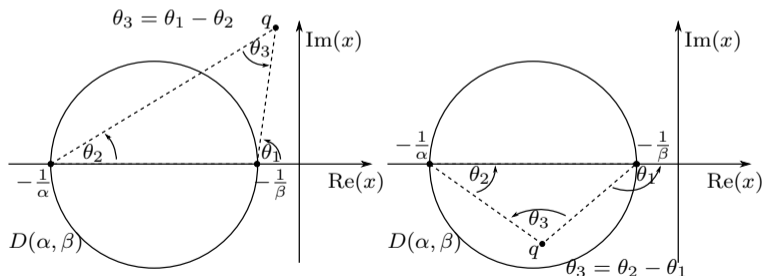
$\rightsquigarrow \operatorname{Re}(H(s)) > 0 \forall \omega \in \mathbb{R} \rightsquigarrow$  Nyquist plot lies entirely inside the disc.

Note that:

- Since the Nyquist plot cannot leave the disc, it cannot encircle the point  $-1/\alpha$
- $G(s)$  cannot have any right-half plane poles, (i.e.,  $G(s)$  is Hurwitz)

Absolute stability:

The Nyquist plot lies entirely inside the disc.



## Circle Criterion (6), (Theorem)

### Theorem (Circle Criterion)

Suppose  $(A, b, c)$  is a minimal realization of  $G(s)$  and  $\psi(t, y)$  satisfies the sector condition

$$\alpha y^2 \leq y\psi(t, y) \leq \beta y^2$$

globally. *Then the system is absolutely stable if:*

- 1  $\alpha = 0 < \beta$ , the Nyquist plot is to the right of the line  $\operatorname{Re}(s) = -\frac{1}{\beta}$ , (i.e., to the right of  $D(0, \beta)$ ) and  $G(s)$  is Hurwitz;
- 2  $0 < \alpha < \beta$ , the Nyquist plot does not enter the disk  $D(\alpha, \beta)$ , and encircles it in the counter-clockwise direction as many times,  $N$ , as there are right-half plane poles of  $G(s)$ ; or
- 3  $\alpha < 0 < \beta$ , the Nyquist plot lies in the interior of the disk  $D(\alpha, \beta)$ , and  $G(s)$  is Hurwitz.



## Circle Criterion (6), (Theorem)

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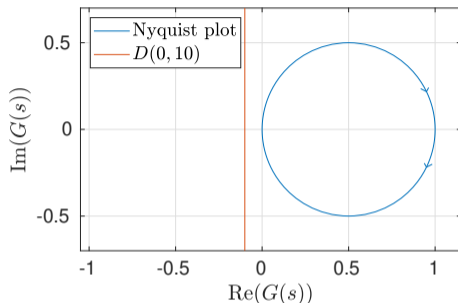
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- 3  $\alpha < 0 < \beta$ , the Nyquist plot lies in the interior of the disk  $D(\alpha, \beta)$ , and  $G(s)$  is Hurwitz.

**Example:** Consider

$$G(s) = \frac{1}{s+1} \quad \text{with pole } s = -1$$

( $G(s)$  is Hurwitz, i.e., three items are potentially applicable.)



- **Item 1** ( $\alpha = 0$  and  $\beta = 10$ ): the Nyquist plot is to the right of the line  $D(0, 10) \rightsquigarrow G(s)$  is absolutely stable

## Circle Criterion (6), (Theorem)

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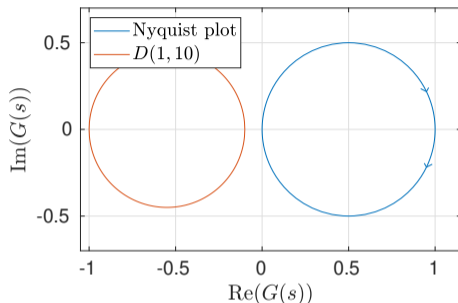
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## Circle Criterion (6), (Theorem)

### Theorem (Circle Criterion)

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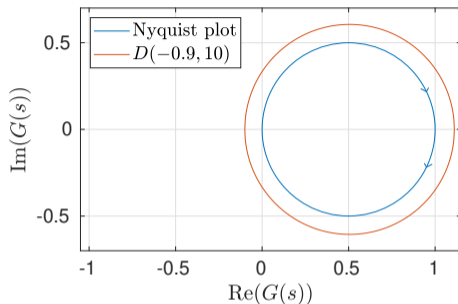
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- 2  $0 < \alpha < \beta$ , the Nyquist plot does not enter the disk  $D(\alpha, \beta)$ , and encircles it in the counter-clockwise direction as many times,  $N$ , as there are right-half plane poles of  $G(s)$ ; or
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**Example:** Consider

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- **Item 2** ( $\alpha = 1$  and  $\beta = 10$ ): the Nyquist plot is outside the disc  $D(1, 10)$  and encircles it zero times  $\rightsquigarrow G(s)$  is absolutely stable
- **Item 3** ( $\alpha = 1$  and  $\beta = 10$ ): the Nyquist is inside the disc  $D(-0.9, 10) \rightsquigarrow G(s)$  is absolutely stable

## Circle Criterion (7), (Examples)

### Theorem (Circle Criterion)

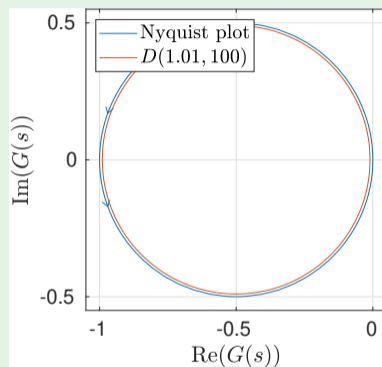
Suppose  $(A, b, c)$  is a minimal realization of  $G(s)$  and  $\psi(t, y)$  satisfies the sector condition

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globally. Then the system is absolutely stable if:

- 1  $\alpha = 0 < \beta$ , the Nyquist plot is to the right of the line  $\text{Re}(s) = -\frac{1}{\beta}$ , (i.e., to the right of  $D(0, \beta)$ ) and  $G(s)$  is Hurwitz;
- 2  $0 < \alpha < \beta$ , the Nyquist plot does not enter the disk  $D(\alpha, \beta)$ , and encircles it in the counter-clockwise direction as many times,  $N$ , as there are right-half plane poles of  $G(s)$ ; or
- 3  $\alpha < 0 < \beta$ , the Nyquist plot lies in the interior of the disk  $D(\alpha, \beta)$ , and  $G(s)$  is Hurwitz.

Example (Consider  $G(s) = \frac{1}{s-1}$ )



- $G(s)$  not Hurwitz; one pole in the right-half plane
- $D(1.01, 100)$  encircles the Nyquist plot exactly once in the counter-clockwise direction
- Absolute stability follows (Item 2)

## Circle Criterion (8), (Examples)

### Theorem (Circle Criterion)

Suppose  $(A, b, c)$  is a minimal realization of  $G(s)$  and  $\psi(t, y)$  satisfies the sector condition

$$\alpha y^2 \leq y\psi(t, y) \leq \beta y^2$$

globally. Then the system is absolutely stable if:

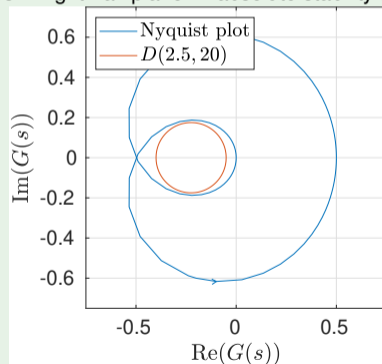
- 1  $\alpha = 0 < \beta$ , the Nyquist plot is to the right of the line  $\text{Re}(s) = -\frac{1}{\beta}$ , (i.e., to the right of  $D(0, \beta)$ ) and  $G(s)$  is Hurwitz;
- 2  $0 < \alpha < \beta$ , the Nyquist plot does not enter the disk  $D(\alpha, \beta)$ , and encircles it in the counter-clockwise direction as many times,  $N$ , as there are right-half plane poles of  $G(s)$ ; or
- 3  $\alpha < 0 < \beta$ , the Nyquist plot lies in the interior of the disk  $D(\alpha, \beta)$ , and  $G(s)$  is Hurwitz.

### Example

Consider the transfer function

$$G(s) = \frac{s+1}{s^2-2s+2} = \frac{s+1}{(s-1+j)(s-1-j)}$$

Two poles in right-half plane  $\rightsquigarrow$  absolute stability (Item 2)

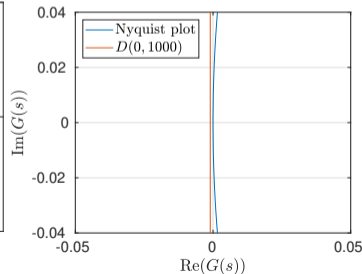
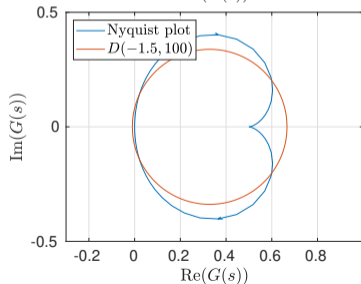
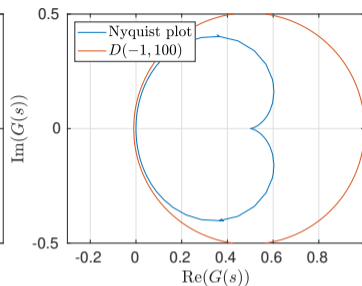
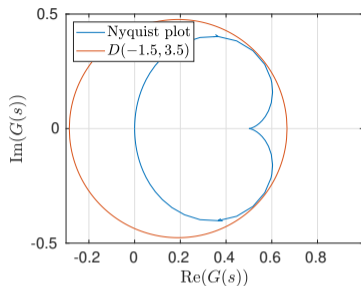


## Circle Criterion (9), (Examples)

Consider the transfer function

$$G(s) = \frac{s+1}{s^2+2s+2}$$
$$= \frac{s+1}{(s+1+j)(s+1-j)}$$

- $G(s)$  is Hurwitz
- Item 3: Absolute stability for  $(\alpha, \beta) = (-1.5, 3.5)$  and for  $(\alpha, \beta) = (-1, 100)$  but not for  $(\alpha, \beta) = (-1.5, 100)$ .
- Item 1:  $\beta$  can be selected arbitrarily large if  $\alpha = 0$ .



# Popov Criterion

## Theorem (Popov Criterion)

Suppose  $A$  is Hurwitz,  $(A, b)$  is controllable,  $(A, c)$  is observable, and  $\psi(y)$  satisfies the sector condition

$$0 \leq y\psi(y) \leq \beta y^2 \quad (1)$$

for all  $y \in \mathbb{R}$ . Then the Lur'e system with  $G(s) = c(sI - A)^{-1}b$  is absolutely stable if there is an  $\eta \geq 0$  with  $-\frac{1}{\eta}$  not an eigenvalue of  $A$  such that

$$H(s) = 1 + (1 + \eta s)\beta G(s)$$

is strictly positive real.

If  $\psi$  only satisfies the sector condition (1) for  $y \in \Omega \subset \mathbb{R}$ , then the system is absolutely stable with a finite domain.

Note that:

- $G(s)$  needs to be Hurwitz
- $\alpha = 0$
- $\psi$  is memoryless, i.e.,  $\psi(t, y) = \psi(y)$

## Proof.

Assume there exists  $\eta \geq 0$  such that  $-\frac{1}{\eta}$  is not an eigenvalue of  $A$  and  $H(s)$  is strictly positive real. Then

$$\begin{aligned} H(s) &= 1 + (1 + \eta s)\beta G(s) \\ &= 1 + \eta\beta cb + c(\beta I + \eta\beta A)(sI - A)^{-1}b \\ &= d + \hat{c}(sI - A)^{-1}b \end{aligned}$$

where  $d = 1 + \eta\beta cb$  and  $\hat{c} = c(\beta I + \eta\beta A)$ .

The condition on  $-\frac{1}{\eta}$  ensures that  $\hat{c} \neq 0$  whenever  $c \neq 0$ .

Since  $H(s)$  is strictly positive real by assumption there exist  $P > 0$ ,  $L$ ,  $\varepsilon$  and  $w$  satisfying the KYP-equations. Consider the candidate Lyapunov function

$$V(x) = x^T Px + 2\eta\beta \int_0^y \psi(r) dr.$$

It can be shown that

$$\dot{V}(x) \leq -\varepsilon x^T Px.$$

which shows absolute stability of the system. □

# Popov Criterion

## Theorem (Popov Criterion)

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for all  $y \in \mathbb{R}$ . Then the Lur'e system with  $G(s) = c(sI - A)^{-1}b$  is absolutely stable if there is an  $\eta \geq 0$  with  $-\frac{1}{\eta}$  not an eigenvalue of  $A$  such that

$$H(s) = 1 + (1 + \eta s)\beta G(s)$$

is strictly positive real.

### Development of a graphical interpretation:

- Recall:  $H(s)$  is strictly positive real if and only if  $H(s)$  is Hurwitz and

$$\operatorname{Re}(1 + (1 + j\eta\omega)\beta G(j\omega)) > 0 \quad \text{for all } \omega \in \mathbb{R}.$$

- $H(s)$  has the same poles as  $G(s)$  since  $-1/\eta$  is not an eigenvalue of  $A$ .

- $G(j\omega) \in \mathbb{C}$  can be written as  $\gamma + j\delta = G(j\omega)$  for  $\gamma, \delta \in \mathbb{R}$  for all  $\omega \in \mathbb{R}$ . Then

$$\begin{aligned} \operatorname{Re}(1 + \beta G(j\omega) + j\eta\omega\beta G(j\omega)) &= \operatorname{Re}(1 + \beta(\gamma + j\delta) + j\eta\omega\beta(\gamma + j\delta)) \\ &= \operatorname{Re}(1 + \beta\gamma - \eta\omega\beta\delta + j(\beta\delta + \eta\omega\beta\gamma)) \\ &= 1 + \beta\gamma - \eta\omega\beta\delta \\ &= 1 + \beta \operatorname{Re}(G(j\omega)) - \eta\omega\beta \operatorname{Im}(G(j\omega)). \end{aligned}$$

- Hence,  $\operatorname{Re}(H(j\omega)) > 0 \forall \omega \in \mathbb{R}$  is equivalent to

$$\frac{1}{\beta} + \operatorname{Re}(G(j\omega)) - \eta\omega \operatorname{Im}(G(j\omega)) > 0.$$

- If we plot  $\operatorname{Re}(G(j\omega))$  versus  $\omega \operatorname{Im}(G(j\omega))$ , the above inequality defines a half space on the right side of the line through  $-\frac{1}{\beta}$  of slope  $\frac{1}{\eta}$ .

- Define the line

$$L(\beta, \eta) = \{x \in \mathbb{C} : x = (-\frac{1}{\beta} + j\frac{1}{\eta})w, w \in \mathbb{R}\}$$

depending on  $\beta > 0$  and  $\eta \geq 0$ .

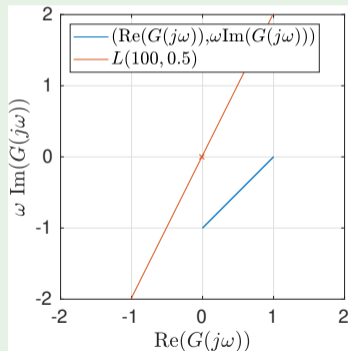
- We refer to a plot of  $\omega \operatorname{Im}(G(j\omega))$  versus  $\operatorname{Re}(G(j\omega))$  including the line  $L(\beta, \eta)$  as a *Popov plot*.



# Popov Criterion

Example (Consider  $G(s) = \frac{1}{s+1}$ )

- $G(s)$  is Hurwitz, i.e., the Popov Criterion is applicable.
- Absolute stability can be concluded for the sector defined through  $\alpha = 0$  and  $\beta = 100$ .



- $G(j\omega) \in \mathbb{C}$  can be written as  $\gamma + j\delta = G(j\omega)$  for  $\gamma, \delta \in \mathbb{R}$  for all  $\omega \in \mathbb{R}$ . Then

$$\begin{aligned} \operatorname{Re}(1 + \beta G(j\omega) + j\eta\omega\beta G(j\omega)) &= \operatorname{Re}(1 + \beta(\gamma + j\delta) + j\eta\omega\beta(\gamma + j\delta)) \\ &= \operatorname{Re}(1 + \beta\gamma - \eta\omega\beta\delta + j(\beta\delta + \eta\omega\beta\gamma)) \\ &= 1 + \beta\gamma - \eta\omega\beta\delta \\ &= 1 + \beta \operatorname{Re}(G(j\omega)) - \eta\omega\beta \operatorname{Im}(G(j\omega)). \end{aligned}$$

- Hence,  $\operatorname{Re}(H(j\omega)) > 0 \forall \omega \in \mathbb{R}$  is equivalent to

$$\frac{1}{\beta} + \operatorname{Re}(G(j\omega)) - \eta\omega \operatorname{Im}(G(j\omega)) > 0.$$

- If we plot  $\operatorname{Re}(G(j\omega))$  versus  $\omega \operatorname{Im}(G(j\omega))$ , the above inequality defines a half space on the right side of the line through  $-\frac{1}{\beta}$  of slope  $\frac{1}{\eta}$ .
- Define the line

$$L(\beta, \eta) = \{x \in \mathbb{C} : x = (-\frac{1}{\beta} + j\frac{1}{\eta})w, w \in \mathbb{R}\}$$

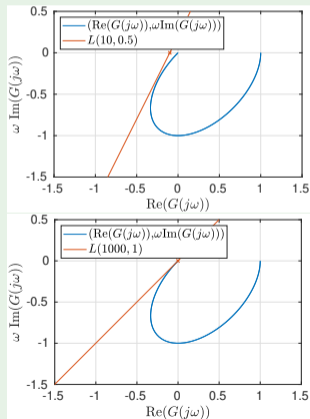
depending on  $\beta > 0$  and  $\eta \geq 0$ .

- We refer to a plot of  $\omega \operatorname{Im}(G(j\omega))$  versus  $\operatorname{Re}(G(j\omega))$  including the line  $L(\beta, \eta)$  as a *Popov plot*.

# Popov Criterion

Example (Consider  $G(s) = \frac{1}{s^2+s+1}$ )

- Two (complex) poles in the open left-half plane.
- Absolute stability can be concluded for different values of  $\beta$  (and  $\alpha = 0$ )



- $G(j\omega) \in \mathbb{C}$  can be written as  $\gamma + j\delta = G(j\omega)$  for  $\gamma, \delta \in \mathbb{R}$  for all  $\omega \in \mathbb{R}$ . Then

$$\begin{aligned} \text{Re}(1 + \beta G(j\omega) + j\eta\omega\beta G(j\omega)) &= \text{Re}(1 + \beta(\gamma + j\delta) + j\eta\omega\beta(\gamma + j\delta)) \\ &= \text{Re}(1 + \beta\gamma - \eta\omega\beta\delta + j(\beta\delta + \eta\omega\beta\gamma)) \\ &= 1 + \beta\gamma - \eta\omega\beta\delta \\ &= 1 + \beta \text{Re}(G(j\omega)) - \eta\omega\beta \text{Im}(G(j\omega)). \end{aligned}$$

- Hence,  $\text{Re}(H(j\omega)) > 0 \forall \omega \in \mathbb{R}$  is equivalent to

$$\frac{1}{\beta} + \text{Re}(G(j\omega)) - \eta\omega \text{Im}(G(j\omega)) > 0.$$

- If we plot  $\text{Re}(G(j\omega))$  versus  $\omega \text{Im}(G(j\omega))$ , the above inequality defines a half space on the right side of the line through  $-\frac{1}{\beta}$  of slope  $\frac{1}{\eta}$ .

- Define the line

$$L(\beta, \eta) = \{x \in \mathbb{C} : x = (-\frac{1}{\beta} + j\frac{1}{\eta})w, w \in \mathbb{R}\}$$

depending on  $\beta > 0$  and  $\eta \geq 0$ .

- We refer to a plot of  $\omega \text{Im}(G(j\omega))$  versus  $\text{Re}(G(j\omega))$  including the line  $L(\beta, \eta)$  as a *Popov plot*.

# Circle versus Popov Criterion

## The Circle Criterion:

- The single-input single-output case discussed here, can be extended to multi-input multi-output systems.
- The Circle Criterion allows for time-varying nonlinearities.
- Checkable conditions for  $\alpha \neq 0$  derived from

$$\frac{1 + \beta G(s)}{1 + \alpha G(s)}.$$

- If  $\eta = 0$  and  $\alpha = 0$  the Circle Criterion and the Popov Criterion are equivalent.

## The Popov Criterion

- Can be extended to the multi-input multi-output settings, but appears to require more structure in the interconnection structure of the input-output behavior and types of nonlinearities that can be accommodated.
- Only applicable to time-invariant nonlinearities.
- Reasonable to assume  $G(s)$  that is Hurwitz and take  $\alpha = 0$  (due to the factor)
- If  $\eta = 0$  and  $\alpha = 0$  the Circle Criterion and the Popov Criterion are equivalent.
- The freedom to choose  $\eta \geq 0$  can provide a less conservative results.
- The assumption  $G(s)$  Hurwitz and  $\alpha = 0$  can be accomplished through an appropriate loop transform.

**Note that:** The Circle Criterion and the Popov Criterion define sufficient conditions.

# Introduction to Nonlinear Control

Stability, control design, and estimation

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## Part I:

### Chapter 6: Absolute Stability

6.1 A Commonly Ignored Design Issue

6.2 Historical Perspective on the Lur'e Problem

6.3 Sufficient Conditions for Absolute Stability



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