

Introduction to Nonlinear Control

Stability, control design, and estimation

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Part II:

Chapter 8: LMI Based Controller and Antiwindup Designs

8.1 \mathcal{L}_2 -gain optimization for linear systems

8.2 Systems with Saturation

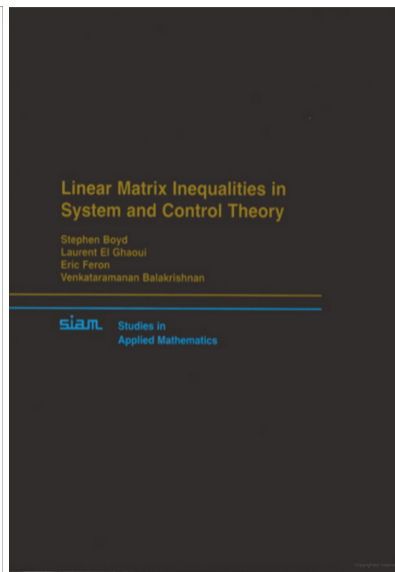
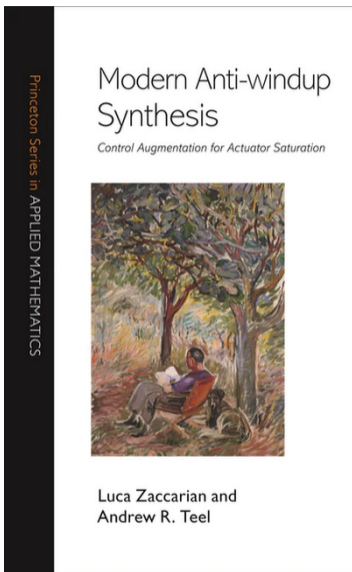
8.3 Regional Analysis

8.4 Antiwindup Synthesis



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LMI Based Controller and Antiwindup Designs



LMI Based Controller and Antiwindup Designs

1 \mathcal{L}_2 -Gain Optimization for Linear Systems

- Asymptotic Stability and \mathcal{L}_2 -Gain Optimization
- Feedback Synthesis

2 Systems with Saturation

- LMI-Based Saturated Linear State Feedback Design
- Global Asymptotic Stability Analysis
- \mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

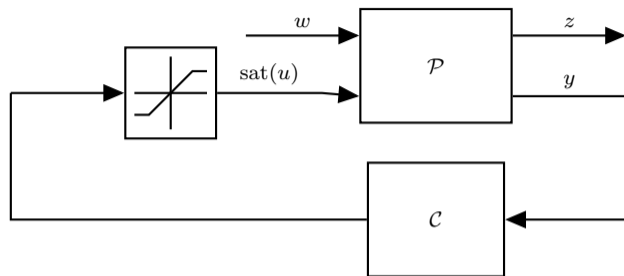
3 Regional Analysis

- Local Asymptotic Stability
- \mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

4 Antiwindup Synthesis

- Global Antiwindup Synthesis
- Regional Antiwindup Synthesis

LMI Based Controller and Antiwindup Designs



Plant & Controller:

$$\mathcal{P} : \begin{cases} \dot{x}_p &= A_p x_p + B_p \text{sat}(u) + B_w w \\ y &= C_{p,y} x_p + D_{p,y} w \\ z &= C_{p,z} x_p + D_{p,z} w \end{cases}$$

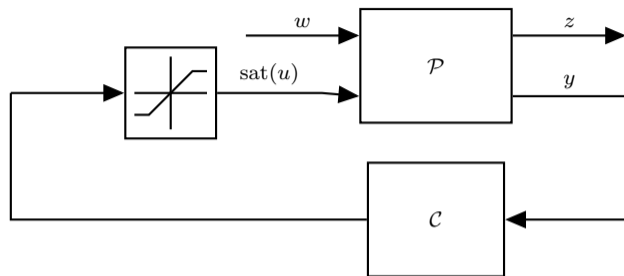
$$\mathcal{C} : \begin{cases} \dot{x}_c &= A_c x_c + B_c y \\ u &= C_c x_c + D_{c,y} y \end{cases}$$

Compact representation: $(x = [x_p^T, x_c^T]^T \in \mathbb{R}^n)$

$$\begin{bmatrix} A & B & E \\ C & D & F \\ K & L & G \end{bmatrix} = \begin{bmatrix} A_p + B_p D_{c,y} C_{p,y} & B_p C_c & -B_p & B_p D_{c,y} D_{p,y} + B_w \\ B_c C_{p,y} & A_c & 0 & B_c D_{p,y} \\ C_{p,z} & 0 & 0 & D_{p,z} \\ D_{c,y} C_{p,y} & C_c & 0 & D_{c,y} D_{p,y} \end{bmatrix}$$

$$\begin{aligned} \dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u) \end{aligned}$$

LMI Based Controller and Antiwindup Designs



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Note that:

- The dynamics on the right is more general than the diagram (L, D)
- The system is nonlinear due to the saturation

- If $L \neq 0$, then $u = L(u - \text{sat}(u)) + \mu$ defines an algebraic loop/equation
- ↔ Existence and uniqueness of a solution is not automatically satisfied

Well-posedness of algebraic loops

System of interest:

$$\begin{aligned}\dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u)\end{aligned}$$

Algebraic loop: ($\mu = Kx + Gw$)

$$u = L(u - \text{sat}(u)) + \mu$$

Definition (Well-posed algebraic loop)

For $L \in \mathbb{R}^{n_u \times n_u}$ consider the algebraic equation. The algebraic equation is well-posed if it admits a unique solution for all $\mu \in \mathbb{R}^{n_u}$ and if $\mu \mapsto u(\mu)$ is Lipschitz continuous.

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Lemma (A sufficient condition)

Consider the algebraic loop for $L \in \mathbb{R}^{n_u \times n_u}$ and $u, \mu \in \mathbb{R}^{n_u}$. If there exists a positive definite matrix $W \in \mathcal{S}_{>0}^{n_u}$ satisfying the matrix inequality

$$\frac{1}{\|W\|} (L^T W + W L - 2W) < 0,$$

then the algebraic loop is well-posed.

Note that

- the factor $\frac{1}{\|W\|}$ is not necessary, but it indicates how far the inequality is from being violated (\rightsquigarrow robustness margin)

Section 1

\mathcal{L}_2 -Gain Optimization for Linear Systems

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Recall: $\dot{x} = Ax$

- The origin is exponentially stable
- For $Q > 0$ there exists $P > 0$ satisfying

$$A^T P + PA = -Q$$

- Idea of the Lyapunov equation is that $V(x) = x^T P x$ is a Lyapunov function

$$\dot{V}(x) = x^T (A^T P + PA)x = -x^T Q x < 0, \quad x \neq 0.$$

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We note that:

- The inequality of the decrease is important not the equality of the Lyapunov equation

↔ For given A , consider the LMI

$$0 < P$$

$$A^T P + PA < 0$$

instead of the Lyapunov equation

- Advantage: Q is a degree of freedom
- “Optimal” Q and P can be obtained

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LMI (as convex optimization problem):

$$\begin{aligned} & \min_{P, k} k \\ \text{subject to} & \quad 0 < k \\ & \quad 0 < P - \alpha I \\ & \quad 0 > P - (k + \alpha)I \\ & \quad 0 > A^T P + PA. \end{aligned}$$

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Here:

- $\alpha > 0$ to ensure that P is not arbitrarily small
- Third constraint to ensure that P is not arbitrarily large

Toolboxes in Matlab:

- CVX, SOSTOOLS, YALMIP

Approximation: ($\varepsilon > 0$)

$$\begin{aligned} \min_{P, k} \quad & k \\ \text{subject to} \quad & 0 \leq k \\ & 0 \leq P - \alpha I - \varepsilon I \\ & 0 \geq P - (k + \alpha)I + \varepsilon I \\ & 0 \geq A^T P + PA + \varepsilon I \end{aligned}$$

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization

Consider:

$$\dot{x} = Ax + Ew$$

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0-GAS: (internal stability)

- If 0 is globally asymptotically stable for $w = 0$, then the system is called 0-GAS
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Recall: For A Hurwitz, $Q = -2I$, $P > 0$ solution of the Lyap. equation, $V(x) = x^T Px$, we have derived

$$\dot{V}(x) \leq -x^T x + \gamma^2 w^T w, \quad \gamma = \|PE\|$$

Rearranging terms and integrating (with $x(0) = 0$) yields

$$\begin{aligned} \|x\|_{\mathcal{L}_2[0,t]}^2 &\leq \int_0^t x(\tau)^T x(\tau) d\tau + V(x(t)) \\ &\leq \gamma^2 \int_0^t w(\tau)^T w(\tau) d\tau = \gamma^2 \|w\|_{\mathcal{L}_2[0,t]}^2. \end{aligned}$$

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Slight modification: Suppose we can find $P > 0$, so that

$$\begin{aligned}\dot{V}(x) &= x^T (A^T P + PA)x + 2x^T P E w \\ &< -\gamma \left(\frac{1}{\gamma^2} z^T z - w^T w \right), \quad \forall (x, w) \neq 0\end{aligned}$$

Then we can show that this guarantees

- 0-GAS (since $\dot{V}(x) < 0 \quad \forall x \neq 0$)
- an \mathcal{L}_2 -gain bound of $\gamma > 0$ from w to output z ; i.e.,

$$\|z\|_{\mathcal{L}_2[0,t]} \leq \gamma \|w\|_{\mathcal{L}_2[0,t]}$$

The bound again follows by integrating (and $x(0) = 0$):

$$\frac{1}{\gamma} \int_0^t z^T(\tau) z(\tau) d\tau + V(x(t)) \leq \gamma \int_0^t w^T(\tau) w(\tau) d\tau$$

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↪ Can we compute $P > 0$ and $\gamma > 0$ by solving an LMI?

Starting point:

$$x^T (A^T P + PA)x + 2x^T P E w + \frac{1}{\gamma} z^T z - \gamma w^T w < 0$$

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization

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$$x^T(A^T P + PA + \frac{1}{\gamma}C^T C)x + 2x^T(PE + \frac{1}{\gamma}C^T F)w + \gamma w^T F^T Fw < 0$$

$$\begin{bmatrix} x \\ w \end{bmatrix}^T \left(\begin{bmatrix} A^T P + PA & PE \\ E^T P & -\gamma I \end{bmatrix} + \frac{1}{\gamma} \begin{bmatrix} C^T \\ F^T \end{bmatrix} \begin{bmatrix} C & F \end{bmatrix} \right) \begin{bmatrix} x \\ w \end{bmatrix} < 0$$

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization

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In terms of definite matrices ($0 < P$ and):

$$\begin{bmatrix} A^T P + PA & PE \\ E^T P & -\gamma I \end{bmatrix} + \frac{1}{\gamma} \begin{bmatrix} C^T \\ F^T \end{bmatrix} \begin{bmatrix} C & F \end{bmatrix} < 0 \quad (1)$$

Note that:

- For $\gamma > 0$, fixed we know how to solve the LMI to obtain P
- However, we would like to minimize $\gamma > 0$
- The inequality is not linear in γ

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization

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Lemma (Schur Complement)

Let $Q \in S^r$ and $R \in S^q$ for $r, q \in \mathbb{N}$ and let $S \in \mathbb{R}^{r \times q}$. Then the matrix condition

$$\begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} < 0$$

is equivalent to the matrix conditions

$$R < 0$$

$$Q - SR^{-1}S^T < 0.$$

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization

Starting point:

$$x^T (A^T P + PA)x + 2x^T P E w + \frac{1}{\gamma} z^T z - \gamma w^T w < 0$$

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In terms of definite matrices ($0 < P$ and):

$$\begin{bmatrix} A^T P + PA & PE \\ E^T P & -\gamma I \end{bmatrix} + \frac{1}{\gamma} \begin{bmatrix} C^T \\ F^T \end{bmatrix} \begin{bmatrix} C & F \end{bmatrix} < 0 \quad (1)$$

Here, take $R = -\gamma$, $S = \begin{bmatrix} C & F \end{bmatrix}$ and Q as the leftmost matrix.

Then, (1) is equivalent to

$$\left[\begin{array}{cc|c} A^T P + PA & PE & C^T \\ E^T P & -\gamma I & F^T \\ \hline C & F & -\gamma I \end{array} \right] < 0$$

Note that:

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$$\begin{aligned} R &< 0 \\ Q - SR^{-1}S^T &< 0. \end{aligned}$$

Asymptotic Stability and \mathcal{L}_2 -Gain Optimization (2)

Overall optimization problem:

$$\begin{aligned} & \min_{P, \gamma} \quad \gamma \\ \text{subject to} \quad & 0 < P \quad \text{symmetric} \\ & 0 < \gamma \\ & 0 > \left[\begin{array}{cc|c} A^T P + PA & PE & C^T \\ E^T P & -\gamma I & F^T \\ \hline C & F & -\gamma I \end{array} \right]. \end{aligned}$$

Notation:

$$\text{He } X = X + X^T$$

Reducing redundancy:

$$\text{He} \left[\begin{array}{cc|c} PA & PE & 0 \\ 0 & -\frac{\gamma}{2} I & 0 \\ \hline C & F & -\frac{\gamma}{2} I \end{array} \right] < 0$$

Note that:

- The information that P is symmetric is redundant
- The constraint $0 < \gamma$ is redundant ($R = -\gamma$)
- Don't forget the factor $\frac{1}{2}$ on the diagonal

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Asymptotic Stability and \mathcal{L}_2 -Gain Optimization (Example)

Consider:

$$\dot{x} = \begin{bmatrix} -1 & -2 & 2 \\ 1 & -2 & 1 \\ 3 & -2 & -2 \end{bmatrix} x + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w$$
$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w.$$

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$$\dot{x} = \begin{bmatrix} -1 & -2 & 2 \\ 1 & -2 & 1 \\ 3 & -2 & -2 \end{bmatrix} x + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w$$
$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w.$$

Solution of the Lyapunov equation with $Q = -2I$:

$$P = \begin{bmatrix} 3.10 & -3.20 & 1.77 \\ -3.20 & 5.50 & -1.80 \\ 1.77 & -1.80 & 1.37 \end{bmatrix}.$$

Estimation (using Young's inequality):

$$\begin{aligned} \|z\|_{\mathcal{L}_2[0,t]}^2 &= \|Cx + Fw\|_{\mathcal{L}_2[0,t]}^2 \\ &= \left(\int_0^t x^T C C x + 2x^T C^T F w + w^T F^T F w \, d\tau \right)^2 \\ &\leq \left(\int_0^t 2x^T C C x + 2w^T F^T F w \, d\tau \right)^2 \\ &\leq \left(\int_0^t 2\lambda_{\max}(C^T C) x^T x + 2\lambda_{\max}(F^T F) w^T w \, d\tau \right)^2 \\ &= 2\lambda_{\max}(C^T C) \|x\|_{\mathcal{L}_2[0,t]}^2 + 2\lambda_{\max}(F^T F) \|w\|_{\mathcal{L}_2[0,t]}^2 \\ &\leq 2 \left(\lambda_{\max}(C^T C) \|PE\|^2 + \lambda_{\max}(F^T F) \right) \|w\|_{\mathcal{L}_2[0,t]}^2 \\ &= 2711 \cdot \|w\|_{\mathcal{L}_2[0,t]}^2. \end{aligned}$$

Hence, $\gamma = \sqrt{2711} = 52.07$

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$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w.$$

Alternatively,

$$\gamma = 7.43$$

$$P = \begin{bmatrix} 4.38 & -0.22 & -4.12 \\ -0.22 & 0.32 & -0.02 \\ -4.12 & -0.02 & 4.18 \end{bmatrix}.$$

is returned as the solution of the LMI

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$$P = \begin{bmatrix} 3.10 & -3.20 & 1.77 \\ -3.20 & 5.50 & -1.80 \\ 1.77 & -1.80 & 1.37 \end{bmatrix}.$$

Estimation (using Young's inequality):

$$\begin{aligned} \|z\|_{\mathcal{L}_2[0,t]}^2 &= \|Cx + Fw\|_{\mathcal{L}_2[0,t]}^2 \\ &= \left(\int_0^t x^T C C x + 2x^T C^T F w + w^T F^T F w \, d\tau \right)^2 \\ &\leq \left(\int_0^t 2x^T C C x + 2w^T F^T F w \, d\tau \right)^2 \\ &\leq \left(\int_0^t 2\lambda_{\max}(C^T C) x^T x + 2\lambda_{\max}(F^T F) w^T w \, d\tau \right)^2 \\ &= 2\lambda_{\max}(C^T C) \|x\|_{\mathcal{L}_2[0,t]}^2 + 2\lambda_{\max}(F^T F) \|w\|_{\mathcal{L}_2[0,t]}^2 \\ &\leq 2 \left(\lambda_{\max}(C^T C) \|PE\|^2 + \lambda_{\max}(F^T F) \right) \|w\|_{\mathcal{L}_2[0,t]}^2 \\ &= 2711 \cdot \|w\|_{\mathcal{L}_2[0,t]}^2. \end{aligned}$$

Hence, $\gamma = \sqrt{2711} = 52.07$

Feedback Synthesis

Consider: (\rightsquigarrow Design static state feedback K)

$$\dot{x} = Ax + Bu + Ew$$

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$$\dot{x} = (A + BK)x + Ew$$

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Same approach as before: ($P > 0$, $V(x) = x^T Px$)

$$\begin{aligned} \dot{V}(x) &= x^T \left((A + BK)^T P + P(A + BK) \right) x + 2x^T PEw \\ &< -\gamma \left(\frac{1}{\gamma^2} z^T z - w^T w \right) \quad \forall (x, w) \neq 0 \end{aligned}$$

(Unknowns: P , K , γ)

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$$\begin{aligned}0 &> (A + BK)^T P + P(A + BK) \\ &= A^T P + PA + K^T B^T P + PBK\end{aligned}$$

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Define $\Lambda = P^{-1}$: (left and right multiplication with Λ)

$$\begin{aligned} \Lambda A^T P \Lambda + \Lambda P A \Lambda + \Lambda K^T B^T P \Lambda + \Lambda P B K \Lambda \\ = \Lambda A^T + A \Lambda + \Lambda (BK)^T + BK \Lambda \\ = \text{He}(\Lambda A + BK \Lambda) \end{aligned}$$

Define $X = K \Lambda$. Then

$$\text{He}(\Lambda A + BK \Lambda) = \text{He}(\Lambda A + BX)$$

is linear in the unknowns $\Lambda = P^{-1}$ and $X = K \Lambda$

The condition

$$\text{He}(\Lambda A + BX) < 0$$

guarantees that $A + BK$ is Hurwitz (with Lyapunov function $V(x) = x^T Px$)

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STEP 2: \mathcal{L}_2 -gain optimization

Recall the condition (for known K)

$$\text{He} \left[\begin{array}{cc|c} P(A + BK) & PE & 0 \\ 0 & -\frac{\gamma}{2} I & 0 \\ \hline C + DK & F & -\frac{\gamma}{2} I \end{array} \right] < 0$$

Feedback Synthesis (2)

Recall the condition:

$$\text{He} \left[\begin{array}{cc|c} P(A+BK) & PE & 0 \\ 0 & -\frac{\gamma}{2}I & 0 \\ \hline C+DK & F & -\frac{\gamma}{2}I \end{array} \right] < 0$$

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Define $\Lambda = P^{-1}$ (left and right multiplication):

$$\begin{bmatrix} \Lambda & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \text{He} \left[\begin{array}{cc|c} P(A+BK) & PE & 0 \\ 0 & -\frac{\gamma}{2}I & 0 \\ \hline C+DK & F & -\frac{\gamma}{2}I \end{array} \right] \begin{bmatrix} \Lambda & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} = \text{He} \left[\begin{array}{cc|c} \Lambda A + BK\Lambda & E & 0 \\ 0 & -\frac{\gamma}{2}I & 0 \\ \hline C\Lambda + DK\Lambda & F & -\frac{\gamma}{2}I \end{array} \right] < 0$$

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Recall the condition:

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Again, define $X = K\Lambda$:

$$\text{He} \left[\begin{array}{cc|c} \Lambda A + BX & E & 0 \\ 0 & -\frac{\gamma}{2}I & 0 \\ \hline C\Lambda + DX & F & -\frac{\gamma}{2}I \end{array} \right] < 0.$$

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Recall the condition:

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Again, define $X = K\Lambda$:

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Corresponding optimization/feasibility problem:

$$\begin{aligned} \min_{\Lambda, X, \gamma} \quad & \gamma \\ \text{subject to} \quad & 0 < \Lambda \quad \text{symmetric} \\ & 0 < \gamma \end{aligned}$$

$$0 > \text{He} \left[\begin{array}{cc|c} (A\Lambda + BX) & E & 0 \\ 0 & -\frac{\gamma}{2}I & 0 \\ \hline C\Lambda + DX & F & -\frac{\gamma}{2}I \end{array} \right]$$

\rightsquigarrow Lyapunov function $V(x) = x^T \Lambda^{-1} x$ and a feedback gain matrix $K = X\Lambda^{-1}$ such that γ is minimal

Feedback Synthesis (Example)

Consider:

$$\dot{x} = \begin{bmatrix} 1 & 2 & -2 \\ -1 & 2 & -1 \\ -3 & 2 & 2 \end{bmatrix} x + \begin{bmatrix} 3 \\ 2 \\ -1 \end{bmatrix} u + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w$$

$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w.$$

Solution of the LMI:

$$\gamma = 8.1910$$

$$K = \begin{bmatrix} -7.32 & 6.64 & 4.62 \end{bmatrix}$$

$$P = \begin{bmatrix} 1.18 & -1.19 & -0.62 \\ -1.19 & 1.33 & 0.71 \\ -0.62 & 0.71 & 0.40 \end{bmatrix}.$$

Eigenvalues of A :

$$\{4, 0.5 \pm 1.32j\}$$

Eigenvalues of $A + BK$:

$$\{-4.78 \pm 0.90j, -0.15\}$$

Eigenvalues of P :

$$\{0.02, 0.07, 2.34\}$$

Section 2

Systems with Saturation

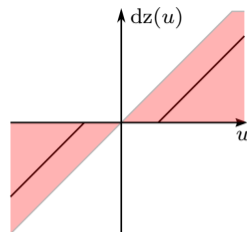
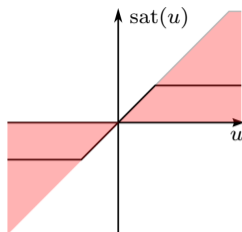
Systems with Saturation

Consider:

$$\begin{aligned}\dot{x} &= Ax + B \text{sat}(u) \\ u &= Kx\end{aligned}$$

Saturation: (we will suppress the limit \bar{u} in the following)

$$\text{sat}(u) \doteq \begin{cases} -1, & u < -1 \\ u, & -1 \leq u \leq 1 \\ 1, & 1 < u. \end{cases}$$
$$\text{sat}_{\bar{u}}(u) \doteq \begin{cases} -\bar{u}, & u < -\bar{u} \\ u, & -\bar{u} \leq u \leq \bar{u} \\ \bar{u}, & \bar{u} < u. \end{cases}$$



Deadzone: ($q = dz(u)$)

$$dz(u) = u - \text{sat}(u) \quad \text{and} \quad dz_{\bar{u}}(u) = u - \text{sat}_{\bar{u}}(u),$$

We assume to have **decentralized saturations**

- i.e., for $u \in \mathbb{R}^{n_u}$ we assume that each input has its own saturation function, possibly with different saturation levels \bar{u}_i on the i^{th} input.

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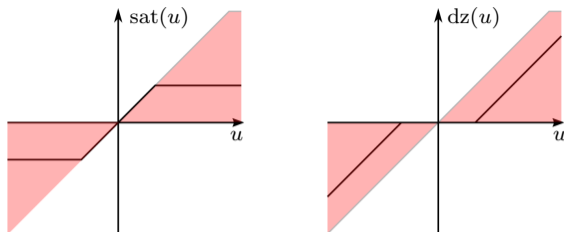
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Note that:

- $u \in \mathbb{R}$ and $q = u - \text{sat}(u)$, satisfies $\text{dz}(u) \text{sat}(u) \geq 0$ or equivalently $q(u - q) \geq 0$.
- In particular $\text{sign}(\text{dz}(u)) = \text{sign}(\text{sat}(u))$ or equivalently $\text{sign}(q) = \text{sign}(u - q)$
- Moreover,

$$wq(u - q) \geq 0 \quad \text{for } w > 0$$

- A vector version: ($W > 0$, diagonal)

$$\text{dz}(u)^T W \text{sat}(u) \geq 0, \quad q^T W (u - q) \geq 0$$

LMI-Based Saturated Linear State Feedback Design

Controller design $u = Kx$: ($q = u - \text{sat}(u)$)

$$\dot{x} = Ax + B \text{sat}(Kx) = (A + BK)x - B \text{dz}(Kx)$$

Consider Lyapunov function: ($P > 0$)

$$V(x) = x^T P x$$

We want: $\dot{V}(x(t)) < 0$ despite the nonlinearity.

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Formally, for $(x, q) = (x, \text{dz}(Kx)) \neq 0$ we want

$$\begin{aligned} \begin{bmatrix} x \\ q \end{bmatrix}^T \text{He} \begin{bmatrix} 0 & 0 \\ -WK & W \end{bmatrix} \begin{bmatrix} x \\ q \end{bmatrix} &= 2q^T W(u - q) \\ &= 2(\text{dz}(Kx))^T W(Kx - \text{dz}(Kx)) \geq 0 \quad \Rightarrow \quad \dot{V}(x) < 0 \end{aligned}$$

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Derivative of the candidate Lyapunov function

$$\begin{aligned} \dot{V}(x) &= x^T ((A + BK)^T P + P(A + BK))x - 2x^T P B q \\ &= x^T (A^T P + P A + K^T B^T P + P B K)x - 2x^T P B \text{dz}(Kx) \\ &= \begin{bmatrix} x \\ q \end{bmatrix}^T \text{He} \begin{bmatrix} P A + P B K & -P B \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x \\ q \end{bmatrix} \end{aligned}$$

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Lemma (S-Lemma or S-Procedure)

Let $M_0, M_1 \in \mathcal{S}^r$, $r \in \mathbb{N}$, and suppose there exists $\zeta^* \in \mathbb{R}^r$ such that $(\zeta^*)^T M_1 \zeta^* > 0$. Then the following statements are equivalent:

- 1 There exists $\tau > 0$ such that $M_0 - \tau M_1 > 0$.
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\rightsquigarrow The inequality guarantees what we want ("1. \Rightarrow 2.")

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In Matrix notation:

$$\begin{bmatrix} x \\ q \end{bmatrix}^T \text{He} \begin{bmatrix} P A + P B K & -P B \\ W K & -W \end{bmatrix} \begin{bmatrix} x \\ q \end{bmatrix} < 0$$

is negative definite.

LMI-Based Saturated Linear State Feedback Design (2)

Controller design $u = Kx$: ($q = u - \text{sat}(u)$)

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If there exists K, P, W such that

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then $u = Kx$ defines a stabilizing control law.

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then $u = Kx$ defines a stabilizing control law.

Multiplying with $\text{diag}(P^{-1}, W^{-1})$:

$$\begin{aligned} 0 &> \begin{bmatrix} P^{-1} & 0 \\ 0 & W^{-1} \end{bmatrix} \text{He} \begin{bmatrix} PA + PBK & -PB \\ WK & -W \end{bmatrix} \begin{bmatrix} P^{-1} & 0 \\ 0 & W^{-1} \end{bmatrix} \\ &= \text{He} \begin{bmatrix} AP^{-1} + BKP^{-1} & -BW^{-1} \\ KP^{-1} & W^{-1} \end{bmatrix} = \text{He} \begin{bmatrix} A\Lambda + BX & -BD \\ X & D \end{bmatrix} \end{aligned}$$

where $\Lambda = P^{-1}$, $X = KP^{-1}$, $D = W^{-1}$

LMI-Based Saturated Linear State Feedback Design (2)

Controller design $u = Kx$: ($q = u - \text{sat}(u)$)

$$\dot{x} = Ax + B \text{sat}(Kx) = (A + BK)x - B \text{dz}(Kx)$$

If there exists K, P, W such that

$$\begin{bmatrix} x \\ q \end{bmatrix}^T \text{He} \begin{bmatrix} PA + PBK & -PB \\ WK & -W \end{bmatrix} \begin{bmatrix} x \\ q \end{bmatrix} < 0$$

then $u = Kx$ defines a stabilizing control law.

Multiplying with $\text{diag}(P^{-1}, W^{-1})$:

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How likely is it that the LMI has a solution?

- If the LMI has a solution then

$$\begin{aligned} 0 > & \begin{bmatrix} I & -B \\ 0 & I \end{bmatrix} \text{He} \begin{bmatrix} A\Lambda + BX & -BD \\ X & -D \end{bmatrix} \begin{bmatrix} I & 0 \\ -B^T & I \end{bmatrix} \\ & = \text{He} \begin{bmatrix} A\Lambda & BD \\ X & -D \end{bmatrix}. \end{aligned}$$

- Then the Schur complement implies that

$$A\Lambda + \Lambda A^T = -Q < 0$$

$$2D + (X^T + BD)Q^{-1}(X + D^T B^T) < 0$$

↪ A is Hurwitz (since $\Lambda > 0$)

↪ The origin is globally asymptotically stable with $K = 0$.

↪ We need local approaches

Global Asymptotic Stability Analysis

Consider

$$\begin{aligned}\dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u)\end{aligned}$$

but we start with $w = 0$.

For K, L given, how do we establish global asymptotic stability of the origin?

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- We want that the sector condition implies a decrease, i.e.,

$$q^T W (u - q) \geq 0 \quad \Rightarrow$$

$$\dot{V}(x) = x^T (A^T P + P A)x + 2x^T P B q < 0, \quad (x, q) \neq 0$$

- S-Procedure: ($W > 0$ diagonal)

$$x^T (A^T P + P A)x + 2x^T P B q + 2q^T W (u - q) < 0$$

- Using the definition of u :

$$\begin{aligned}x^T (A^T P + P A)x + 2x^T P B q \\ + 2q^T W (Kx - Lq - q) < 0\end{aligned}$$

- Corresponding LMI:

$$\text{He} \begin{bmatrix} PA & PB \\ WK & -W + WL \end{bmatrix} < 0.$$

↪ Feasibility (unknowns $P > 0, W > 0$ diagonal) implies global asymptotic stability

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↪ Feasibility (unknowns $P > 0, W > 0$ diagonal) implies global asymptotic stability

- Note that: The Schur complement implies

$$-2W + WL + L^T W < 0$$

Recall: Well-posedness of the algebraic loop

$$u = Kx + L(u - \text{sat}(u)) + Gw$$

Lemma (A sufficient condition)

Consider the algebraic loop for $L \in \mathbb{R}^{n_u \times n_u}$ and $u, \mu \in \mathbb{R}^{n_u}$. If there exists a positive definite matrix $W \in \mathcal{S}_{>0}^{n_u}$ satisfying the matrix inequality

$$\frac{1}{\|W\|} \left(L^T W + W L - 2W \right) < 0,$$

then the algebraic loop is well-posed.

Global Asymptotic Stability Analysis (2)

Consider (with $w = 0$)

$$\begin{aligned}\dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u)\end{aligned}$$

- Corresponding LMI:

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- **Note that:** The Schur complement implies

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$$\frac{1}{\|W\|} (L^T W + WL - 2W) < 0,$$

then the algebraic loop is well-posed.

Introducing a robustness margin: ($\nu \in (0, 1]$)

$$\begin{aligned}x^T (A^T P + PA)x + 2x^T PBq + 2q^T W(Kx - Lq - q) &\leq \\ x^T (A^T P + PA)x + 2x^T PBq + 2q^T W(Kx - Lq - \nu q) &\leq 0\end{aligned}$$

Corresponding LMI:

$$\text{He} \begin{bmatrix} PA & PB \\ WK & -\nu W + WL \end{bmatrix} < 0 \quad (2)$$

The Schur Complement implies:

$$-2\nu W + WL + L^T W < 0.$$

If (2) is satisfied, then

$$\frac{1}{\|W\|} (L^T W + WL - 2W) < -2(1 - \nu) \frac{W}{\|W\|} < 0$$

\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

Consider

$$\begin{aligned} \dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u) \end{aligned}$$

\mathcal{L}_2 -stability problem (for given K, L). We perform the same steps as before

$$q^T W(u - q) \geq 0 \quad \Rightarrow \quad x^T (A^T P + PA)x + 2x^T P(Bq + Ew) < -\gamma \left(\frac{1}{\gamma^2} z^T z - w^T w \right)$$

$$\text{(S-Procedure)} \quad x^T (A^T P + PA)x + 2x^T P(Bq + Ew) + 2q^T W(u - q) + \gamma \left(\frac{1}{\gamma^2} z^T z - w^T w \right) < 0$$

$$\text{(Definition of } u) \quad x^T (A^T P + PA)x + 2x^T P(Bq + Ew) + 2q^T W(Gw + Kx + Lq - q) + \frac{1}{\gamma} z^T z - \gamma w^T w < 0$$

Corresponding matrix notation:

$$\begin{bmatrix} x \\ q \\ w \end{bmatrix}^T \left(\begin{bmatrix} A^T P + PA & PB + K^T W & PE \\ B^T P + WK & -2W + WL + L^T W & WG \\ E^T P & G^T W & -\gamma I \end{bmatrix} + \frac{1}{\gamma} \begin{bmatrix} C^T \\ D^T \\ F^T \end{bmatrix} [C \ D \ F] \right) \begin{bmatrix} x \\ q \\ w \end{bmatrix} < 0.$$

Schur complement

$$\text{He} \left[\begin{array}{ccc|c} PA & PB & PE & 0 \\ WK & -W + WL & WG & 0 \\ 0 & 0 & -\frac{\gamma}{2} I & 0 \\ \hline C & D & F & -\frac{\gamma}{2} I \end{array} \right] < 0.$$

\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization (Optimization Problem & Example)

Overall Optimization problem: ($\nu \in (0, 1]$)

$$\begin{aligned} & \min_{P, W, \gamma} \quad \gamma \\ \text{subject to} \quad & 0 < P \quad \text{symmetric} \\ & 0 < W \quad \text{diagonal} \\ & 0 < \gamma \\ & 0 > \text{He} \left[\begin{array}{ccc|c} PA & PB & PE & 0 \\ WK & -\nu W + WL & WG & 0 \\ 0 & 0 & -\frac{\gamma}{2} I & 0 \\ \hline C & D & F & -\frac{\gamma}{2} I \end{array} \right] \end{aligned}$$

Example:

Consider

$$\dot{x} = \begin{bmatrix} -1 & -2 & 2 \\ 1 & -2 & 1 \\ 3 & -2 & -2 \end{bmatrix} x + \begin{bmatrix} 3 \\ 2 \\ -1 \end{bmatrix} q + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w$$

$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w$$

$$u = \begin{bmatrix} -1 & -2 & 1 \end{bmatrix} x + \begin{bmatrix} 2 & -3 \end{bmatrix} w$$

$$q = u - \text{sat}(u).$$

Solution: $\gamma = 7.8607$ (for $\nu = 1$)

Section 3

Regional Analysis

Regional Analysis

Lemma

Let $u \in \mathbb{R}$ and $q = u - \text{sat}(u)$. For an arbitrary row vector $H \in \mathbb{R}^{1 \times n}$, for all $x \in \mathbb{R}^n$ such that $\text{sat}(Hx) = Hx$, the sector condition

$$(u - q + Hx)q \geq 0 \quad \text{holds.}$$

Proof.

Let $x \in \mathbb{R}^n$ such that $\text{sat}(Hx) = Hx$ is satisfied. Then

$$\begin{aligned}(u - q + Hx)q &= (u - u + \text{sat}(u) + Hx)(u - \text{sat}(u)) \\ &= (\text{sat}(u) + Hx)(u - \text{sat}(u)) \\ &= (\text{sat}(u) + \text{sat}(Hx))(u - \text{sat}(u))\end{aligned}$$

Now, consider two cases:

- 1 $u = \text{sat}(u)$: Then $(\text{sat}(u) + \text{sat}(Hx))(u - \text{sat}(u)) = 0$
- 2 $u \neq \text{sat}(u)$: Both terms on the right have the same sign or are zero. Specifically: (1) if $u > \text{sat}(u)$ then $\text{sat}(u) \geq \text{sat}(Hx)$; (2) if $u = \text{sat}(u)$ the right term vanishes; (3) if $u < \text{sat}(u)$ then $\text{sat}(u) \leq \text{sat}(Hx)$.

□

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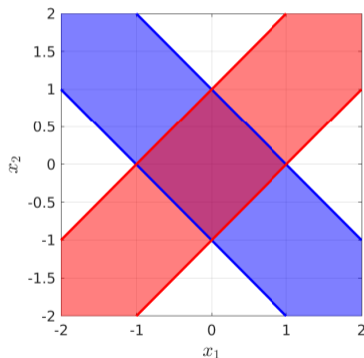
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□

Vector version and positive scaling $W > 0$ diagonal:

$$(u - q + Hx)^T W q \geq 0$$

Visualization of the sector condition:



Domain where $\text{sat}(Hx) = Hx$ is satisfied. Here $H_b = [1, 1]$ (blue) and $H_r = [1, -1]$ (red).

Local Asymptotic Stability

Candidate Lyapunov function:

$$V(x) = x^T P x$$

We want

$$V(x) = x^T P x \leq 1 \quad \Rightarrow \quad \dot{V}(x) < 0, \quad x \neq 0$$

Note that: (\bar{u}_i saturation level)

$$\frac{1}{\bar{u}_i^2} x^T H_i^T H_i x < x^T P x, \quad \forall x \neq 0, \quad \forall i = 1, \dots, n_u$$

implies ($i \in \{1, \dots, n_u\}$)

$$\text{sat}_{\bar{u}_i}(H_i x) = H_i x \quad \forall x \in \{x \in \mathbb{R}^{n_u} \mid x^T P x \leq 1\}$$

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Schur complement: (unknowns P, H_i)

$$0 < \begin{bmatrix} P & H_i^T \\ H_i & \bar{u}_i^2 \end{bmatrix}, \quad i = 1, \dots, n_u$$

Consider (with $w = 0$)

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\rightsquigarrow We can proceed as before.

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$$\dot{V}(x) = x^T (A^T P + PA)x + 2x^T P B q < 0, \quad (x, q) \neq 0$$

and we apply the S-procedure

$$\begin{aligned} &x^T (A^T P + PA)x + 2x^T P B q \\ &+ 2q^T W(Kx - Lq - q + Hx) < 0 \quad \forall (x, q) \neq 0 \end{aligned}$$

and obtain the matrix representation

$$\text{He} \begin{bmatrix} PA & PB \\ WK + WH & -W + WL \end{bmatrix} < 0$$

\rightsquigarrow Not an LMI due to WH

Local Asymptotic Stability (2)

Define set of new unknowns:

$$\Lambda_1 = P^{-1}, \quad \Lambda_2 = W^{-1}, \quad \Gamma = HP^{-1}$$

Note that:

$$\begin{bmatrix} \Gamma_1 \\ \vdots \\ \Gamma_{n_u} \end{bmatrix} = \Gamma = HP^{-1} = \begin{bmatrix} H_1 P^{-1} \\ \vdots \\ H_{n_u} P^{-1} \end{bmatrix}$$

$$\begin{aligned} 0 &< \begin{bmatrix} P^{-1} & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} P & H_i^T \\ H_i & \bar{u}_i^2 \end{bmatrix} \begin{bmatrix} P^{-1} & 0 \\ 0 & 1 \end{bmatrix} \\ &= \begin{bmatrix} P^{-1} & P^{-1}H_i^T \\ H_i P^{-1} & \bar{u}_i^2 \end{bmatrix} = \begin{bmatrix} \Lambda_1 & \Gamma_i^T \\ \Gamma_i & \bar{u}_i^2 \end{bmatrix} \end{aligned}$$

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Corresponding optimization problem:

$$\begin{aligned} \min_{\Lambda_1, \Lambda_2, \Gamma, k} \quad & k \\ \text{subject to} \quad & 0 < \Lambda_1 \quad \text{symmetric} \\ & 0 < \Lambda_2 \quad \text{diagonal} \\ & 0 < k \\ & 0 < kI - \Lambda_1 \\ & 0 < \text{He} \begin{bmatrix} \frac{1}{2}\Lambda_1 & 0 \\ \Gamma_i & \frac{1}{2}\bar{u}_i^2 \end{bmatrix}, \quad i = 1, \dots, n_u \\ & 0 > \text{He} \begin{bmatrix} A\Lambda_1 & B\Lambda_2 \\ K\Lambda_1 + \Gamma & -\Lambda_2 + L\Lambda_2 \end{bmatrix} \end{aligned}$$

- If the the optimization problem is feasible then the origin is locally asymptotically stable
- Estimate of the region of attraction $\{x \in \mathbb{R}^n : x^T P x \leq 1\}$ (with $P = \Lambda_1^{-1}$)
- The smallest eigenvalue of P is maximized:

$$0 < kI - \Lambda_1 \iff \frac{1}{k}I < P$$
- $\nu \in (0, 1)$ can be incorporated

Local Asymptotic Stability (Example)

Consider

$$\begin{aligned}\dot{x} &= Ax + B \operatorname{sat}(u) + Ew \\ z &= Cx + D \operatorname{sat}(u) + Fw \\ u &= Kx.\end{aligned}$$

Using the deadzone operator:

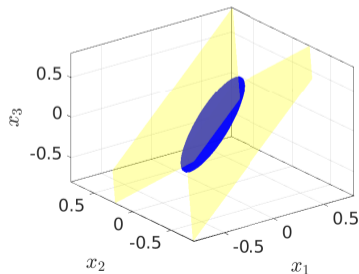
$$\begin{aligned}\dot{x} &= Ax + BKx - Bq + Ew = (A + BK)x - Bq + Ew \\ z &= Cx + DBx - Dq + Fw = (C + DB)x - Dq + Fw \\ u &= Kx \\ q &= u - \operatorname{sat}(u)\end{aligned}$$

We continue with an earlier example (which we have stabilized in the unconstrained case):

$$\begin{aligned}\dot{x} &= \begin{bmatrix} -20.93 & 21.92 & 11.83 \\ -15.62 & 15.28 & 8.22 \\ 4.31 & -4.64 & -2.61 \end{bmatrix} x + \begin{bmatrix} -3 \\ -2 \\ 1 \end{bmatrix} q + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w \\ z &= \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w \\ u &= \begin{bmatrix} -7.31 & 6.64 & 4.61 \end{bmatrix} x + \begin{bmatrix} 2 & -3 \end{bmatrix} w \\ q &= u - \operatorname{sat}(u)\end{aligned}$$

- (For $w = 0$) the system without saturation is globally asymptotically stable
- Saturation level $\bar{u} = 1$, ($|\operatorname{sat}(Hx)| = 1$)
- We obtain $V(x) = x^T P x$ with

$$P = \begin{bmatrix} 21.93 & -22.11 & -10.78 \\ -22.11 & 26.01 & 12.83 \\ -10.78 & 12.83 & 9.43 \end{bmatrix}$$
$$H = [2.75, -1.98, -2.06]$$



\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

We continue with:

$$\begin{aligned} \dot{x} &= Ax + Bq + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + Lq + Gw \\ q &= u - \text{sat}(u) \end{aligned}$$

- Local asymptotic stability: $V(x) = x^T Px$,

$$\dot{V}(x) < 0 \quad \forall x \in \{x \in \mathbb{R}^n \setminus \{0\} \mid V(x) \leq 1\}$$

\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

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- Local \mathcal{L}_2 -stability: For $s > 0$ fixed, find $V(x) = x^T Px$ and $\gamma > 0$ such that

$$\|z\|_{\mathcal{L}_2} \leq \gamma \|w\|_{\mathcal{L}_2} \quad \forall \|w\|_{\mathcal{L}_2} \leq s$$

(and $x(0) = 0$).

- Derive conditions based on

$$\dot{V}(x(t)) \leq w^T w \quad \forall x \in \{x \in \mathbb{R}^n : V(x) \leq s^2\}$$

- Corresponding LMI:

$$0 < \begin{bmatrix} P & H_i^T \\ H_i & \frac{\bar{u}_i^2}{s^2} \end{bmatrix}, \quad i = 1, \dots, n_u.$$

\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization

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- Corresponding LMI:

$$0 < \begin{bmatrix} P & H_i^T \\ H_i & \frac{\bar{u}_i^2}{s^2} \end{bmatrix}, \quad i = 1, \dots, n_u.$$

Overall optimization problem:

$$\begin{aligned} \min_{\Lambda_1, \Lambda_2, \Gamma, \delta} \quad & \delta \\ \text{subject to} \quad & 0 < \Lambda_1 \quad \text{symmetric} \\ & 0 < \Lambda_2 \quad \text{diagonal} \\ & 0 < \delta \end{aligned}$$

$$0 < \text{He} \begin{bmatrix} \frac{1}{2}\Lambda_1 & 0 \\ \Gamma_i & \frac{\bar{u}_i^2}{2s^2} \end{bmatrix} \quad i = 1, \dots, n_u$$

$$0 > \text{He} \begin{bmatrix} A\Lambda_1 & B\Lambda_2 & E & 0 \\ \Gamma + K\Lambda_1 & -\Lambda_2 + L\Lambda_2 & G & 0 \\ 0 & 0 & -\frac{1}{2}I & 0 \\ C\Lambda_1 & D\Lambda_2 & F & -\frac{\delta}{2}I \end{bmatrix}$$

Feasibility for fixed $s > 0$ implies:

- The local \mathcal{L}_2 -bound for $\gamma = \sqrt{\delta}$
- local asymptotic stability for all $x \in \mathbb{R}^n$ contained in the sublevel set $\{x \in \mathbb{R}^n : x^T Px \leq s^2\}$.

\mathcal{L}_2 -Stability and \mathcal{L}_2 -Gain Optimization (Example)

We continue with the example:

$$\dot{x} = Ax + B \text{sat}(u) + Ew$$

$$z = Cx + D \text{sat}(u) + Fw$$

$$u = Kx.$$

Using the deadzone operator:

$$\dot{x} = Ax + BKx - Bq + Ew = (A + BK)x - Bq + Ew$$

$$z = Cx + DBx - Dq + Fw = (C + DB)x - Dq + Fw$$

$$u = Kx$$

$$q = u - \text{sat}(u)$$

We continue with an earlier example (which we have stabilized in the unconstrained case):

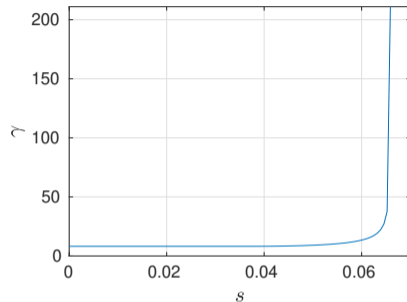
$$\dot{x} = \begin{bmatrix} -20.93 & 21.92 & 11.83 \\ -15.62 & 15.28 & 8.22 \\ 4.31 & -4.64 & -2.61 \end{bmatrix} x + \begin{bmatrix} -3 \\ -2 \\ 1 \end{bmatrix} q + \begin{bmatrix} 2 & -2 \\ 1 & 3 \\ 3 & -2 \end{bmatrix} w$$

$$z = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} -3 & 2 \end{bmatrix} w$$

$$u = \begin{bmatrix} -7.31 & 6.64 & 4.61 \end{bmatrix} x + \begin{bmatrix} 2 & -3 \end{bmatrix} w$$

$$q = u - \text{sat}(u)$$

Optimal \mathcal{L}_2 -gain γ (with respect to s):



Section 4

Antiwindup Synthesis

Antiwindup Synthesis

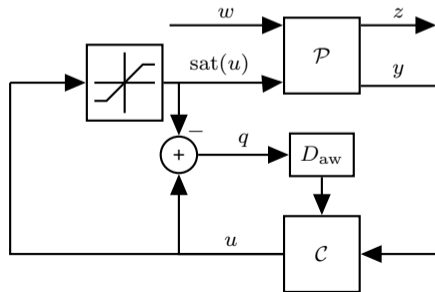
Plant & Controller:

$$\mathcal{P} : \begin{cases} \dot{x}_p &= A_p x_p + B_p \text{sat}(u) + B_w w \\ y &= C_{p,y} x_p + D_{p,y} w \\ z &= C_{p,z} x_p + D_{p,z} w \end{cases}$$

$$\mathcal{C} : \begin{cases} \dot{x}_c &= A_c x_c + B_c y + D_{aw,1} q \\ u &= C_c x_c + D_{c,y} y + D_{aw,2} q \end{cases}$$

Anti-windup injection terms

- $D_{aw,1}$ and $D_{aw,2}$ are to be designed to improve the closed-loop performance.



Updated system dynamics:

$$\begin{aligned} \dot{x} &= Ax + (B + B_{aw} D_{aw})q + Ew \\ z &= Cx + Dq + Fw \\ u &= Kx + (L + L_{aw} D_{aw})q + Gw \\ q &= u - \text{sat}(u) \end{aligned}$$

Design parameter:

$$D_{aw} = \begin{bmatrix} D_{aw,1} \\ D_{aw,2} \end{bmatrix}$$

System/Controller parameter:

$$B_{aw} = \begin{bmatrix} 0 & B_p \\ I_{n_c} & 0 \end{bmatrix}, \quad L_{aw} = [0 \ I_{n_u}].$$

Global Antiwindup Synthesis

Optimization problem:

$$\begin{aligned} & \min_{\Lambda_1, \Lambda_2, X, \gamma} \quad \gamma \\ & \text{subject to} \quad 0 < \Lambda_1 \quad \text{symmetric} \\ & \quad \quad \quad 0 < \Lambda_2 \quad \text{diagonal} \\ & \quad \quad \quad 0 < \gamma \\ & \quad \quad \quad 0 > \text{He} \left[\begin{array}{ccc|c} A\Lambda_1 & B\Lambda_2 + B_{\text{aw}}X & E & 0 \\ K\Lambda_1 & -\nu\Lambda_2 + L\Lambda_2 + L_{\text{aw}}X & G & 0 \\ 0 & 0 & -\frac{\gamma}{2}I & 0 \\ \hline C\Lambda_1 & D & F & -\frac{\gamma}{2}I \end{array} \right] \end{aligned}$$

↪ If the optimization problem is feasible, the antiwindup injection term $D_{\text{aw}} = X\Lambda_2^{-1}$

↪ $\nu \in (0, 1]$ can be used/decreased to obtain an implementable $D_{\text{aw},2}$ (well-posedness of algebraic loop)

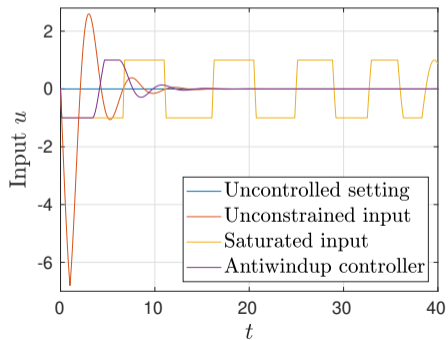
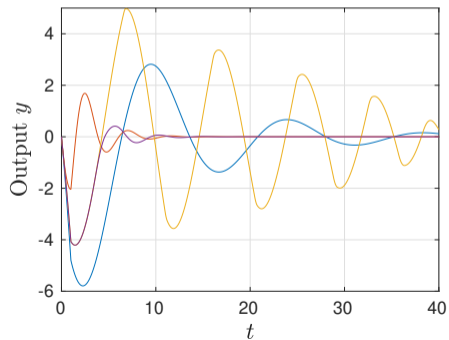
↪ Local analysis can be performed using the same tricks discussed before

Global Antiwindup Synthesis (Example)

Consider the plant/controller defined through the dynamics (subject to the disturbances):

$$\left[\begin{array}{c|c|c} A_p & B_p & B_w \\ \hline C_{p,y} & & D_{p,y} \\ \hline C_{p,z} & & D_{p,z} \end{array} \right] = \left[\begin{array}{cc|c|c} -0.2 & -0.2 & 0.6 & 3 \\ 1 & 0 & 0.4 & 3 \\ \hline -0.4 & -0.9 & & 0 \\ \hline -0.4 & -0.9 & & 0 \end{array} \right], \quad \left[\begin{array}{c|c} A_c & B_c \\ \hline C_c & D_{c,y} \end{array} \right] = \left[\begin{array}{c|c} 0 & 1 \\ \hline 2 & 2 \end{array} \right], \quad w(t) = \begin{cases} 1, & \text{if } t \leq 1 \\ 0, & \text{if } t > 1 \end{cases}$$

$$\rightsquigarrow D_{aw,1} = -127.30, D_{aw,2} = 0.45.$$



Introduction to Nonlinear Control

Stability, control design, and estimation

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Part II:

Chapter 8: LMI Based Controller and Antiwindup Designs

8.1 \mathcal{L}_2 -gain optimization for linear systems

8.2 Systems with Saturation

8.3 Regional Analysis

8.4 Antiwindup Synthesis



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